

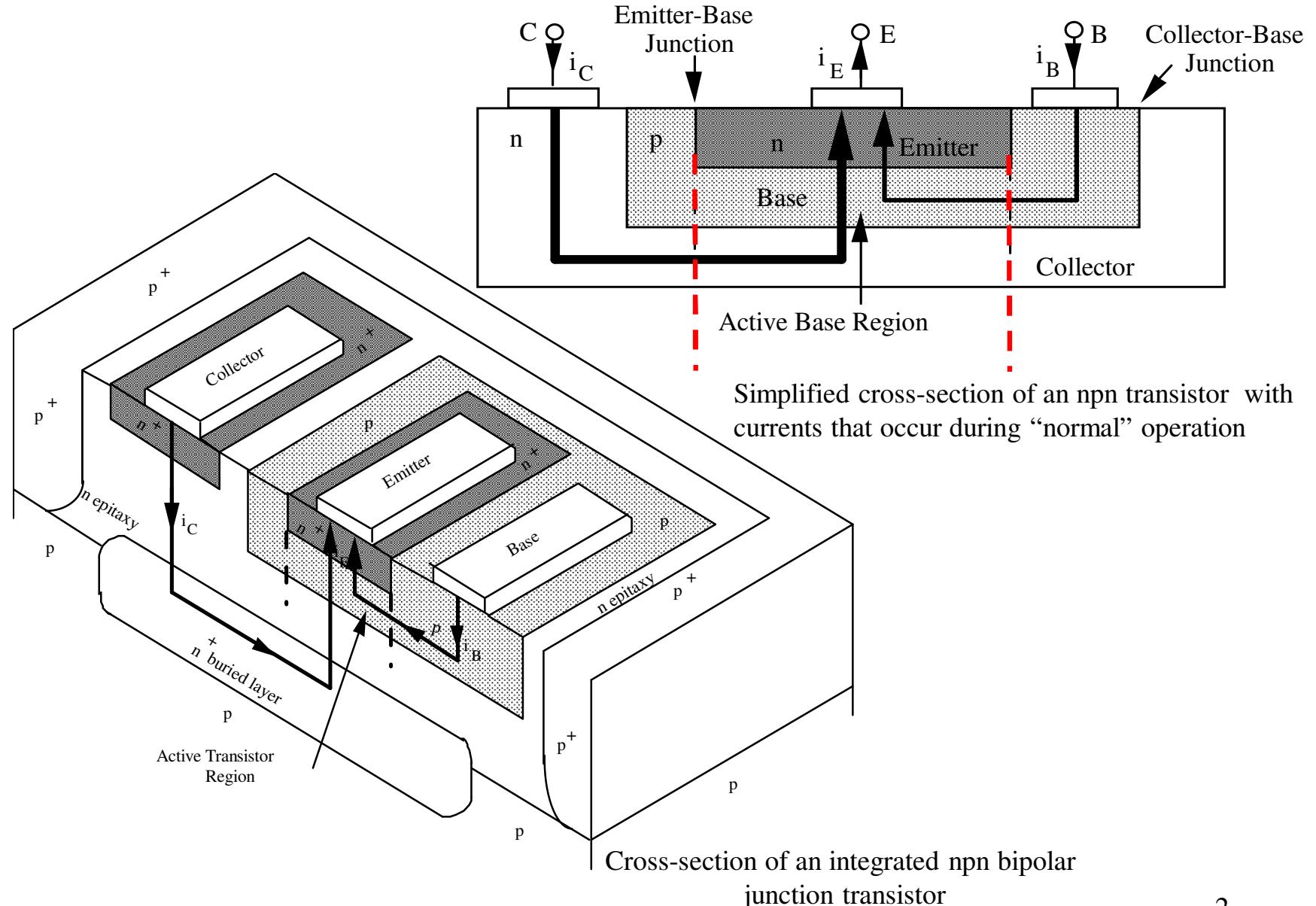
The Bipolar Junction Transistor

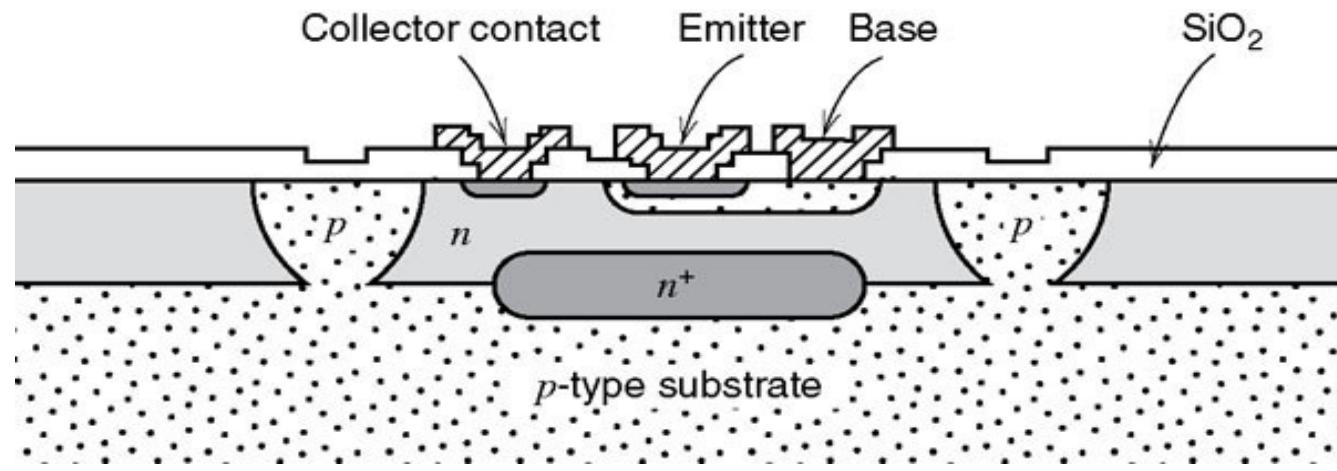
- Physical Structure of the Bipolar Transistor
- Operation of the NPN Transistor in the Active Mode
- Transit Time and Diffusion Capacitance
- Injection Efficiency and Base Transport Factor
- The Ebers-Moll Model of the BJT
- The PNP Transistor
- Graphical Representation of the BJT Characteristics
- The Early Effect

Appendices:

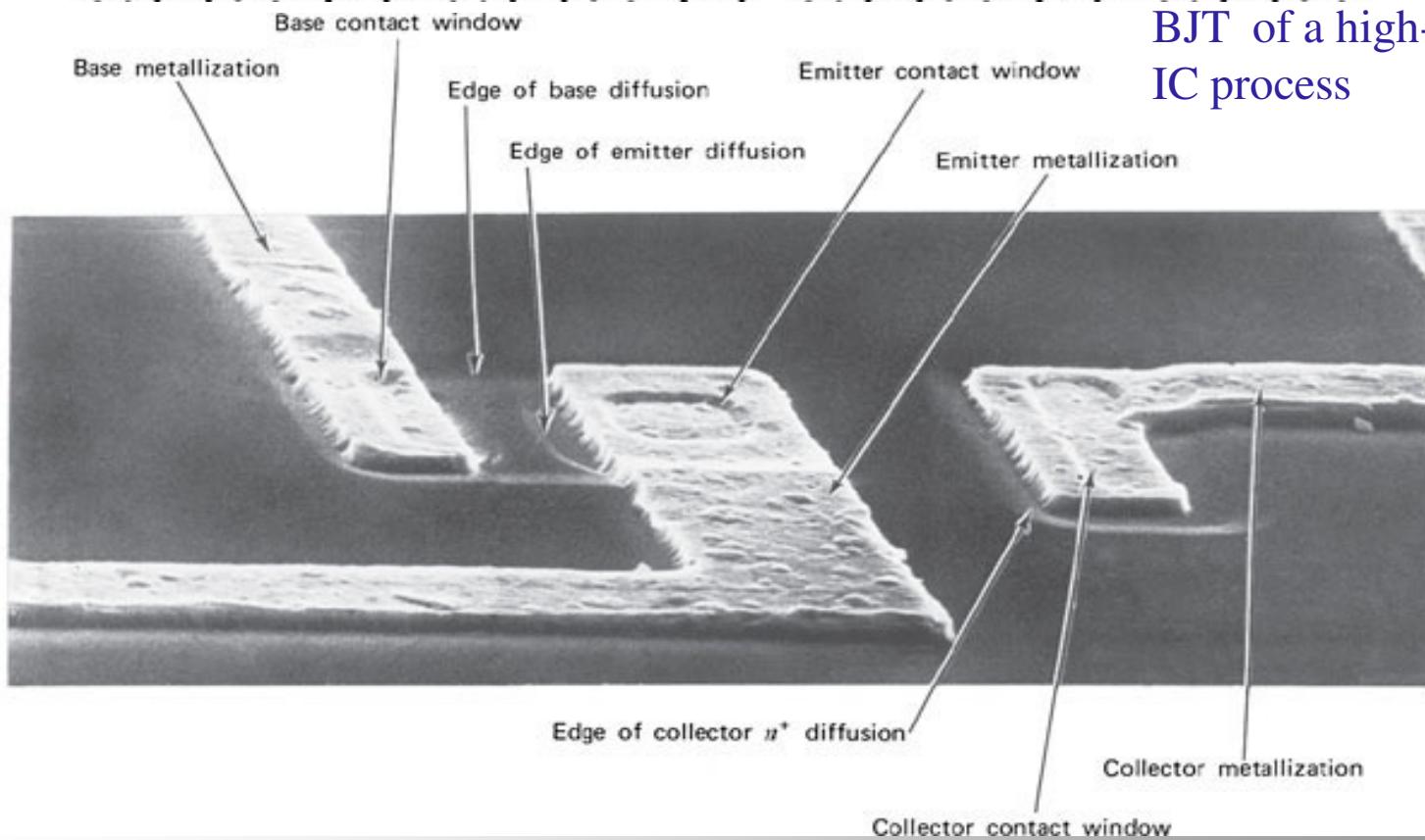
- BJT parameters for simulation

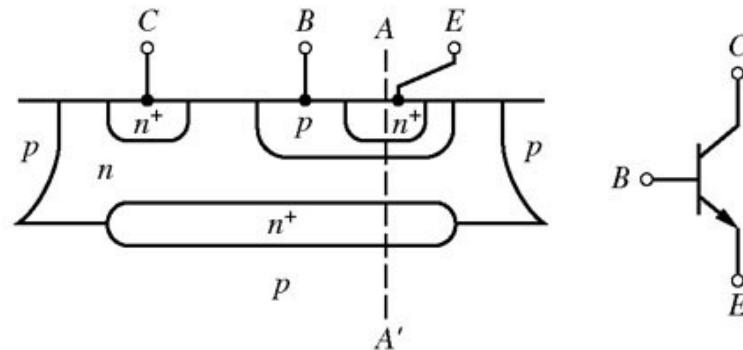
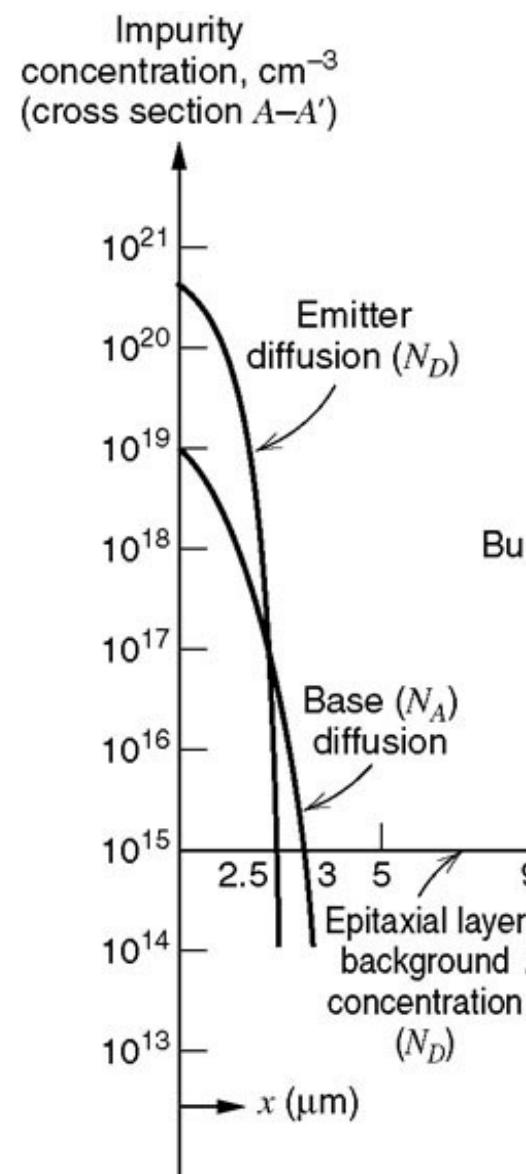
Physical Structure of the BJT



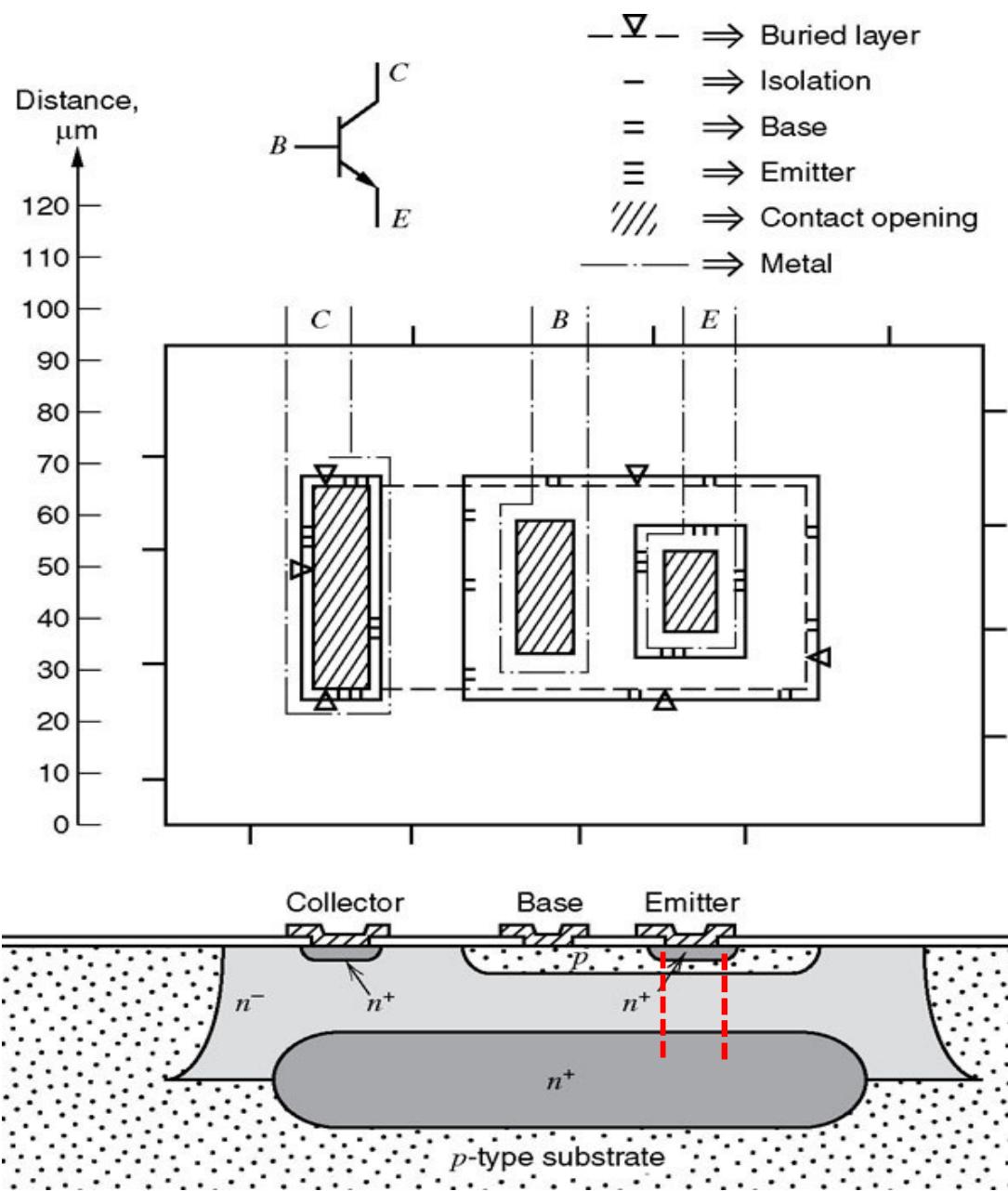


BJT of a high-voltage
IC process

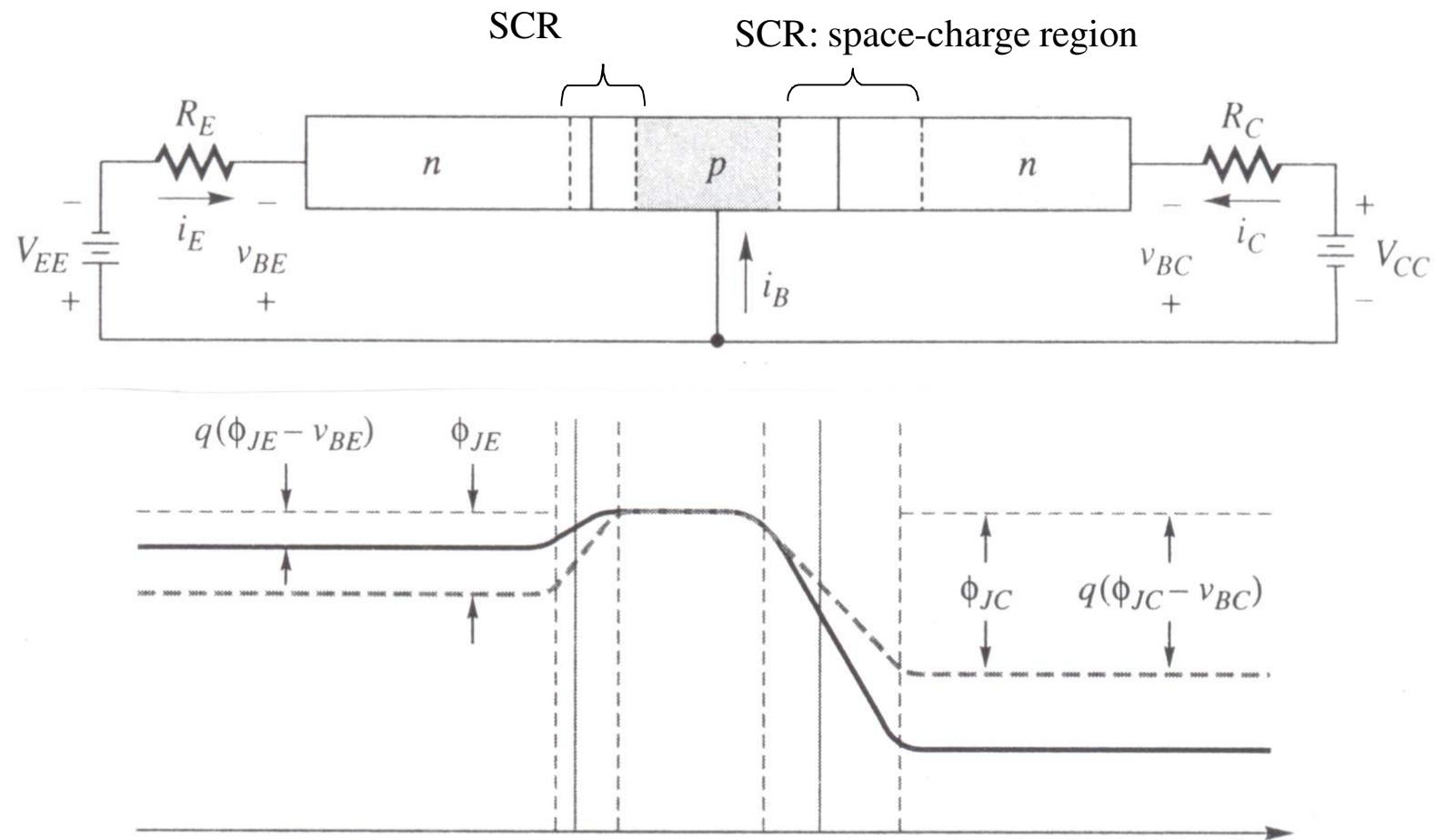




Typical impurity concentration for a monolithic npn transistor in a high-voltage process



Operation of the BJT in the Active Mode



- (a) npn transistor with base-emitter junction forward biased, base-collector junction reverse biased. (b) Potential-energy barriers for electrons.

Operation of the BJT in the Active Mode (*)

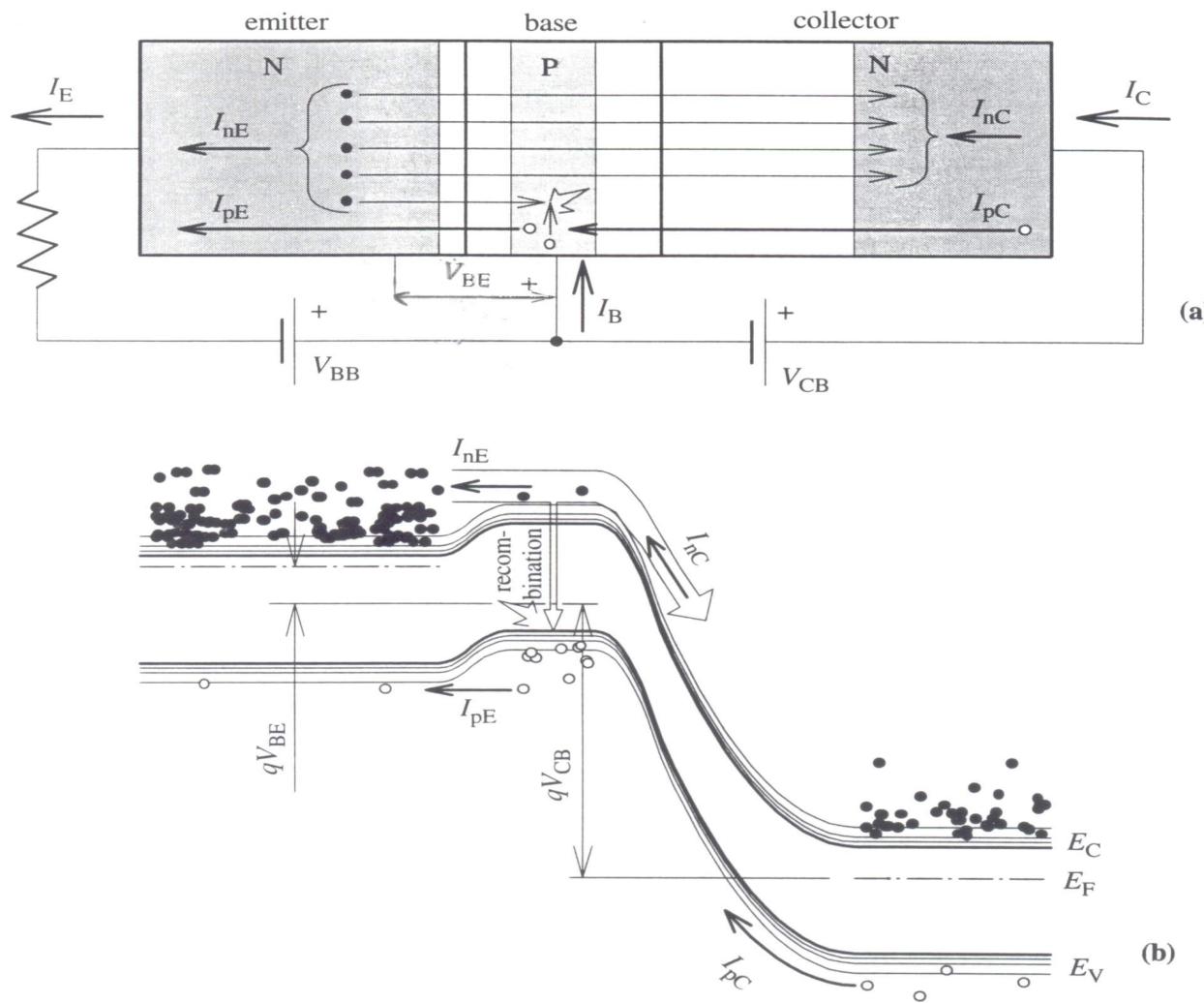
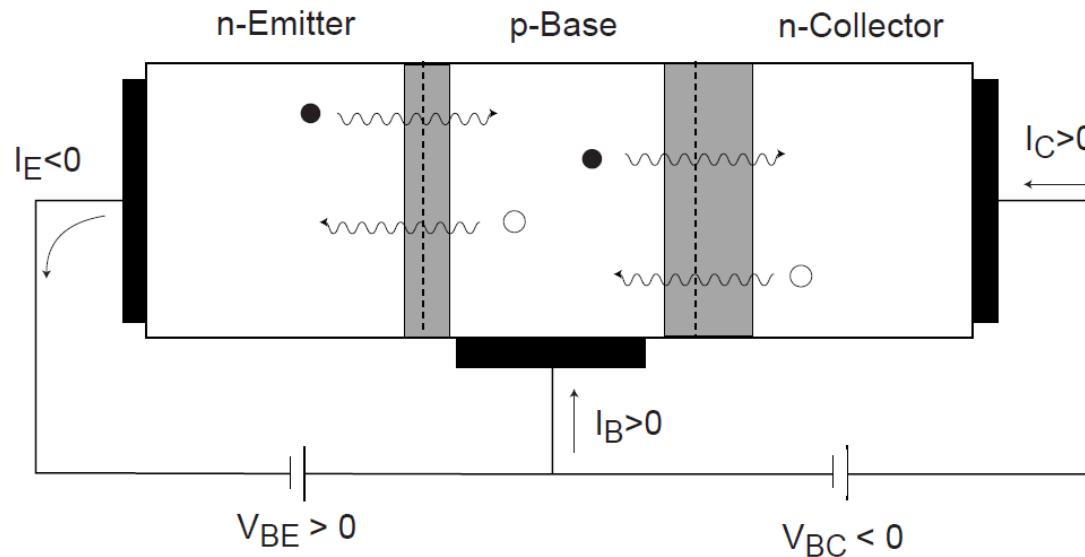


Illustration of BJT currents, using the BJT cross section (a) and energy-band diagram (b).

(*) S. Dimitrijev, Understanding Semiconductor Devices, Oxford, New York, 2000. 7

Basic Operation: forward-active regime



$V_{BE} > 0 \Rightarrow$ injection of electrons from the **Emitter** to the **Base**

injection of holes from the **Base** to the **Emitter**

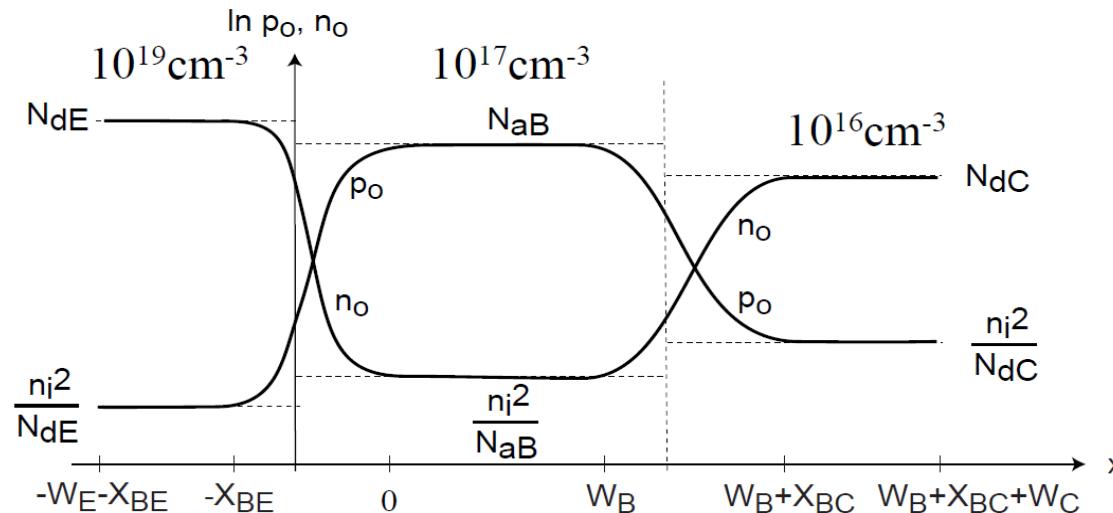
$V_{BC} < 0 \Rightarrow$ extraction of electrons from the **Base** to the **Collector**

extraction of holes from the **Collector** to the **Base**

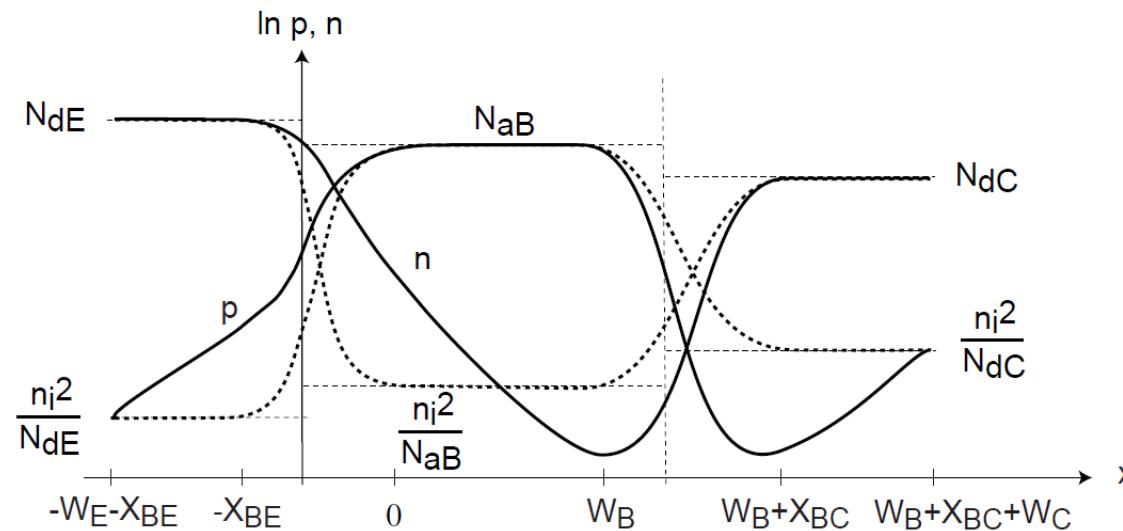
Transistor Effect : electrons injected from the **Emitter** to the **Base**, extracted by the **Collector**

Basic Operation: forward-active regime

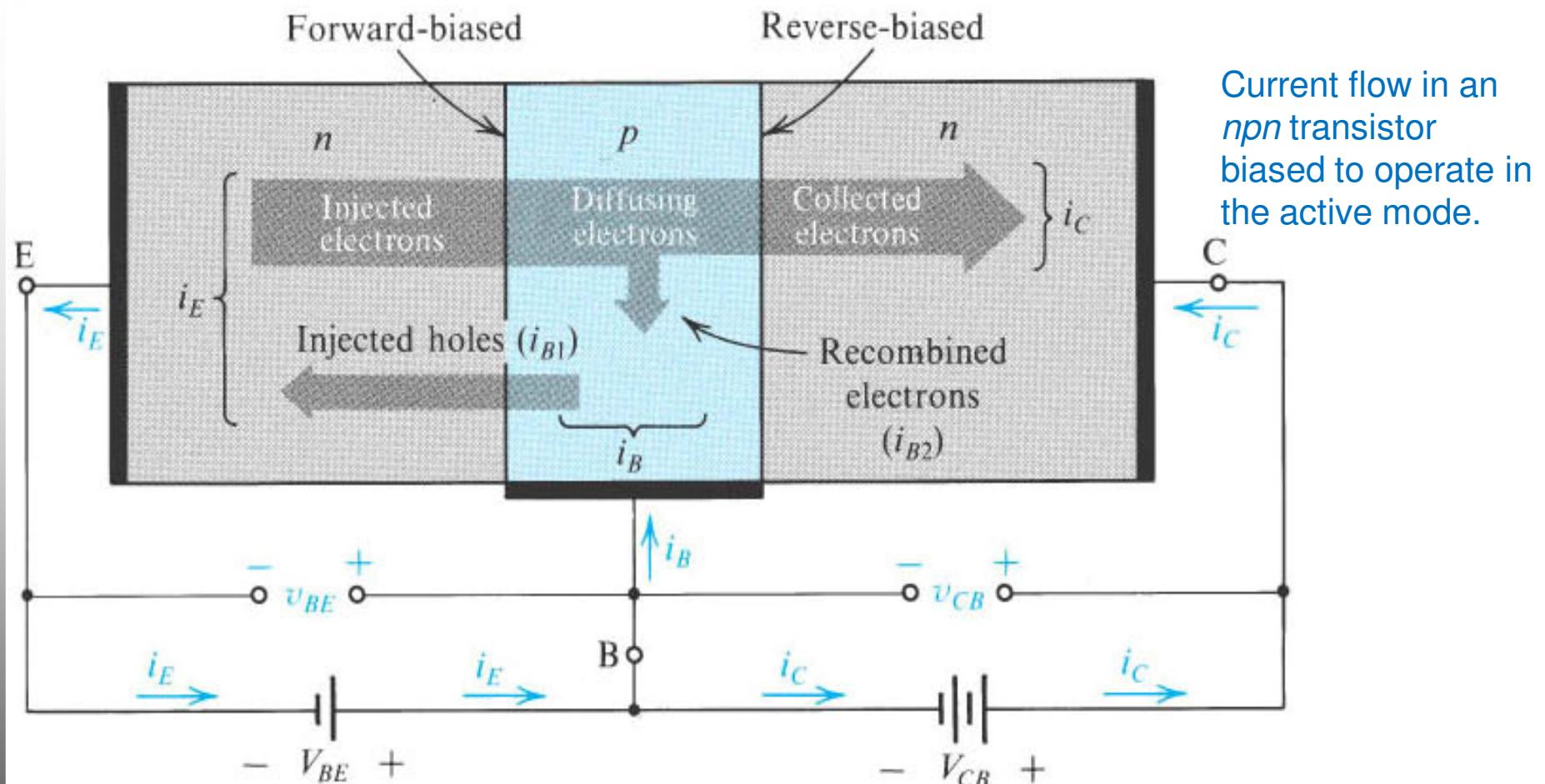
- Carrier profiles in thermal equilibrium:



- Carrier profiles in forward-active regime:



BJT in the Active Mode

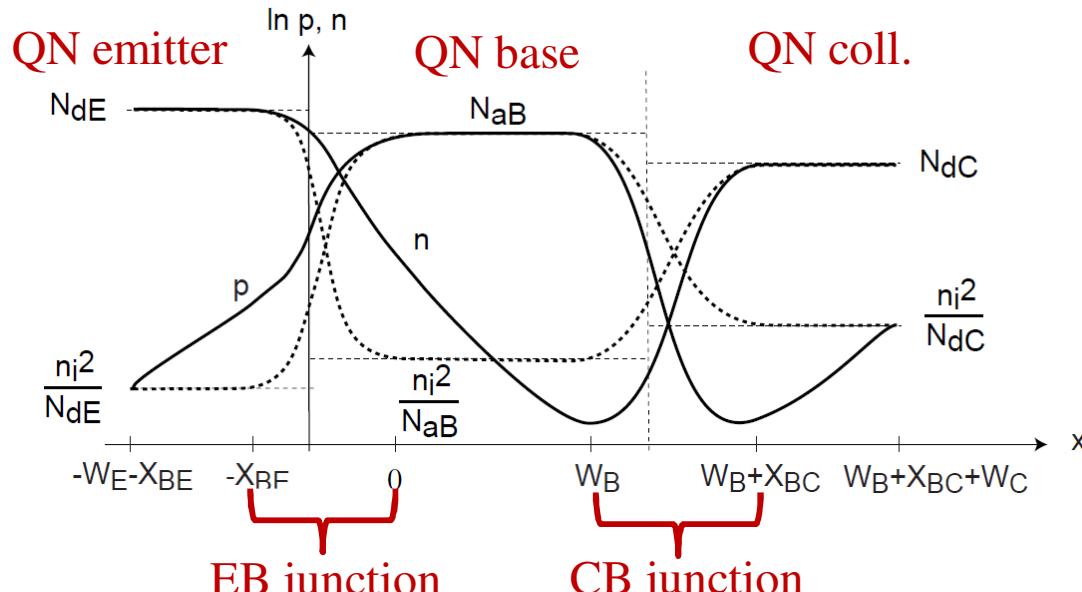


Current flow in an *n*p*n* transistor biased to operate in the active mode.

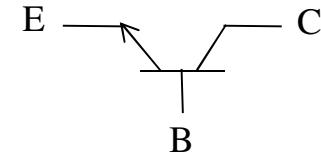
- The electron flux “stream” is greater than the hole flux stream.
- Electrons are supplied by the emitter contact, injected across the base-emitter SCR and diffuse across the base
- Electric field in the base-collector SCR extracts electrons into the collector.
- Holes are supplied by the base contact and diffuse across the emitter.
- The reverse injected holes recombine at the emitter ohmic contact.

BJT in the Active Mode

Carrier profiles in forward-active regime:



$QN = \text{quasi-neutral}$



Hypotheses:

- One-dimensional device
- Ohmic contacts at E, B, and C
- 3 QN regions and 2 depletion regions
- Low injection level
- High emitter efficiency
- Einstein relationship
- Boltzmann statistics valid in the depletion region
- $W_B \ll L_n$ – negligible recombination in the base

The electron (collector) current - npn transistor

One-dimensional device + Einstein relationship

$$\left. \begin{aligned} J_n &= qn\mu_n E + qD_n \frac{dn}{dx} \rightarrow \frac{J_n}{qn\mu_n} - \frac{\phi_t}{n} \frac{dn}{dx} = E \\ J_p &= qp\mu_p E - qD_p \frac{dp}{dx} \rightarrow \frac{J_p}{qp\mu_p} + \frac{\phi_t}{p} \frac{dp}{dx} = E \end{aligned} \right\} \frac{pJ_n}{D_n} - \frac{nJ_p}{D_p} = q \frac{d(pn)}{dx} \quad (1)$$

In the QN base (1) becomes

$$\left. \begin{aligned} \text{Low injection level} \quad p &\equiv N_{aB} \gg n \\ \text{High emitter efficiency} \quad J_n &> J_p \end{aligned} \right\} \frac{pJ_n}{D_n} \approx q \frac{d(pn)}{dx} \quad (2)$$

For negligible recombination $J_n \approx \text{constant}$. Integrating (2) from 0 to W_B results in

$$\begin{aligned} J_n \int_0^{W_B} pdx &\approx qD_n \int_0^{W_B} d(pn) = qD_n \cdot pn \Big|_0^{W_B} = -qD_n n_i^2 \left[e^{\frac{V_{BE}}{\phi_t}} - e^{\frac{V_{BC}}{\phi_t}} \right] \\ \int_0^{W_B} pdx &\approx \int_0^{W_B} N_{aB} dx \rightarrow I_n = AJ_n = -\frac{qAD_n n_i^2}{\int_0^{W_B} N_{aB} dx} \left[e^{\frac{V_{BE}}{\phi_t}} - e^{\frac{V_{BC}}{\phi_t}} \right] = -I_S \left[e^{\frac{V_{BE}}{\phi_t}} - e^{\frac{V_{BC}}{\phi_t}} \right] \end{aligned}$$

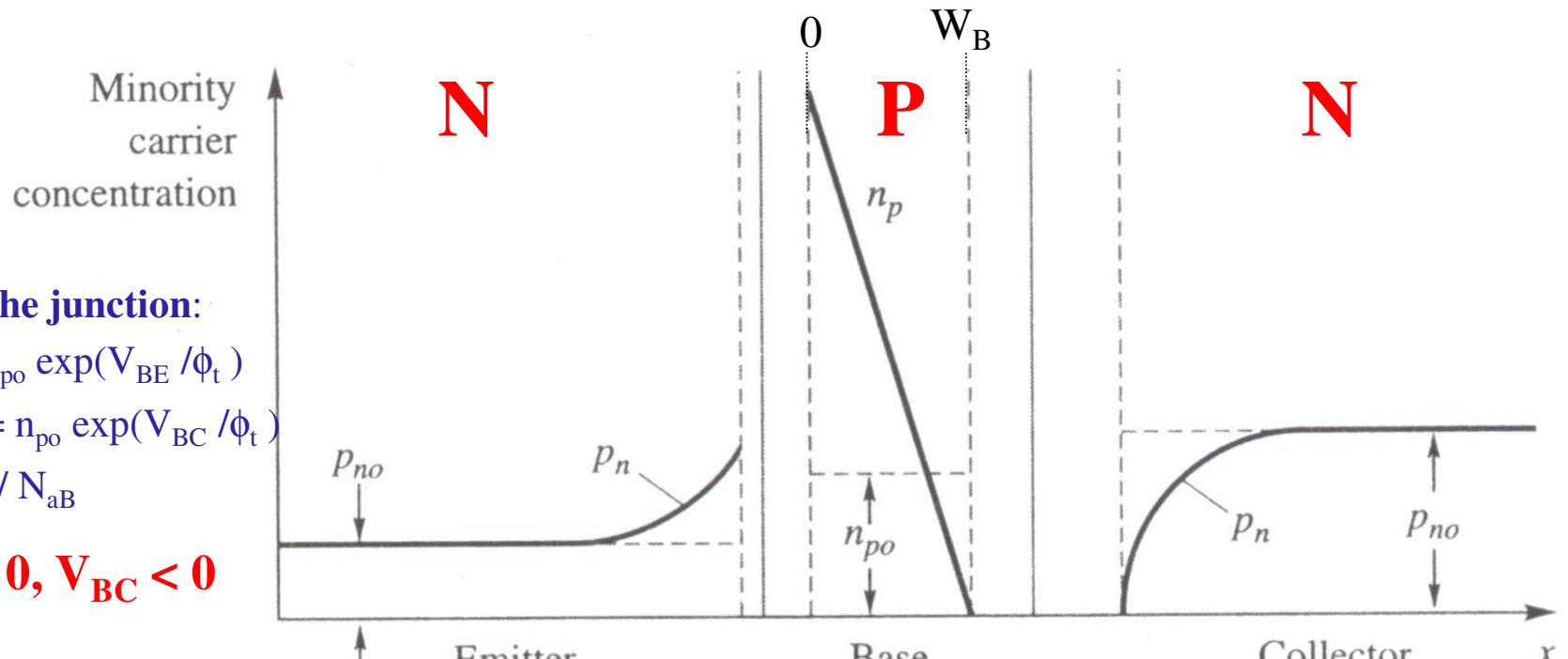
Law of the junction:

$$n_p(0) = n_{po} \exp(V_{BE}/\phi_t)$$

$$n_p(W_B) = n_{po} \exp(V_{BC}/\phi_t)$$

$$n_{po} = n_i^2 / N_{aB}$$

$$V_{BE} > 0, V_{BC} < 0$$



(c) Minority carrier concentrations in the various regions of the transistor.

For constant impurity concentration in the base and low injection level

$$p \equiv N_{aB} \rightarrow \frac{p J_n}{D_n} \equiv q \frac{d(pn)}{dx} \rightarrow \frac{dn}{dx} = \frac{J_n}{q D_n} = \frac{I_n}{q A D_n} \rightarrow \text{Diffusion current}$$

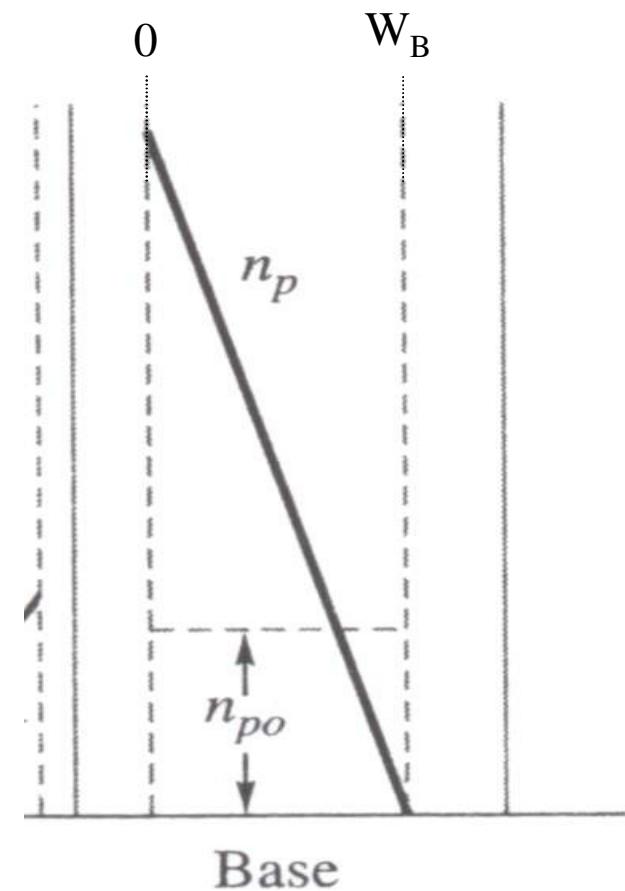
$$I_n = - \frac{q A D_n n_i^2}{W_B} \int_0^{W_B} N_{aB} dx \left[e^{\frac{V_{BE}}{\phi_t}} - e^{\frac{V_{BC}}{\phi_t}} \right] = - \frac{q A D_n n_i^2}{N_{aB} W_B} \left[e^{\frac{V_{BE}}{\phi_t}} - e^{\frac{V_{BC}}{\phi_t}} \right]$$

BJT in the Active Mode

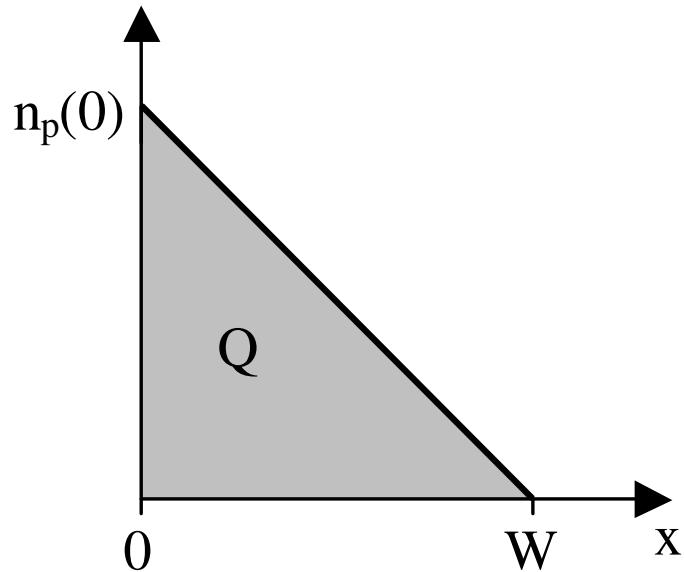
$$V_{BE} > 0, V_{BC} < 0$$

$$n_{p0} = \frac{n_i^2}{N_{aB}}; \quad n_p(0) = n_{p0} e^{\frac{V_{BE}}{\phi_t}}; \quad n_p(W_B) = n_{p0} e^{\frac{V_{BC}}{\phi_t}}$$

$$n_p(x) \equiv n_p(0) \left[1 - \frac{x}{W_B} \right]$$



Forward Transit Time



Excess minority charge stored in the neutral base

Operating frequency

$$f \leq \frac{1}{2\pi\tau_F}$$

$$Q = \frac{q n_p(0) A_E W}{2}$$

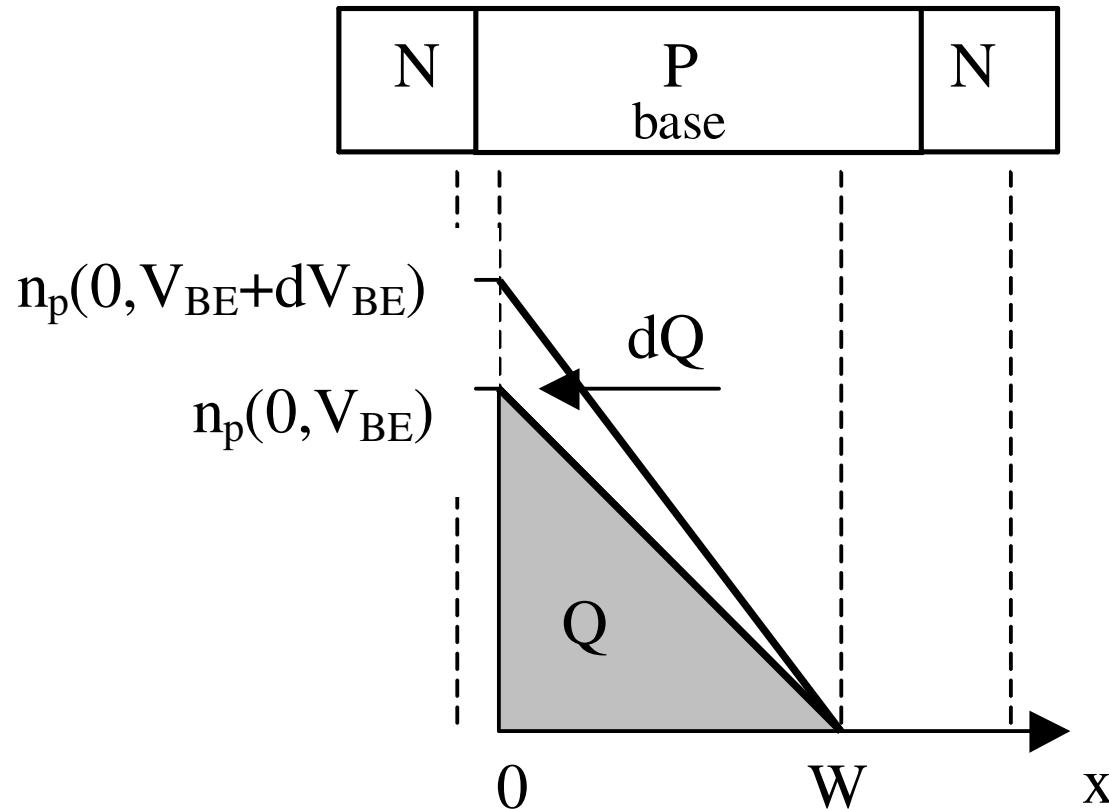
$$\tau_F = \frac{Q}{I_C} = \frac{W^2}{2D_n}$$

If $W = 10^{-4}$ cm ($= 1\mu\text{m}$),
 $D_n = 12.5 \text{ cm}^2/\text{s}$ ($= \mu_n \phi_t$)

$$\tau_F = \frac{10^{-8} \text{ cm}^2}{2 \times 12.5 \text{ cm}^2/\text{s}}$$

$$\boxed{\tau_F = 4 \times 10^{-10} \text{ s}}$$

Diffusion Capacitance (C_D)



$$C_D = \frac{dQ}{dV_{BE}}$$

$$Q = \tau_F I_C$$

$$I_C = I_s e^{\frac{V_{BE}}{\phi_t}}$$

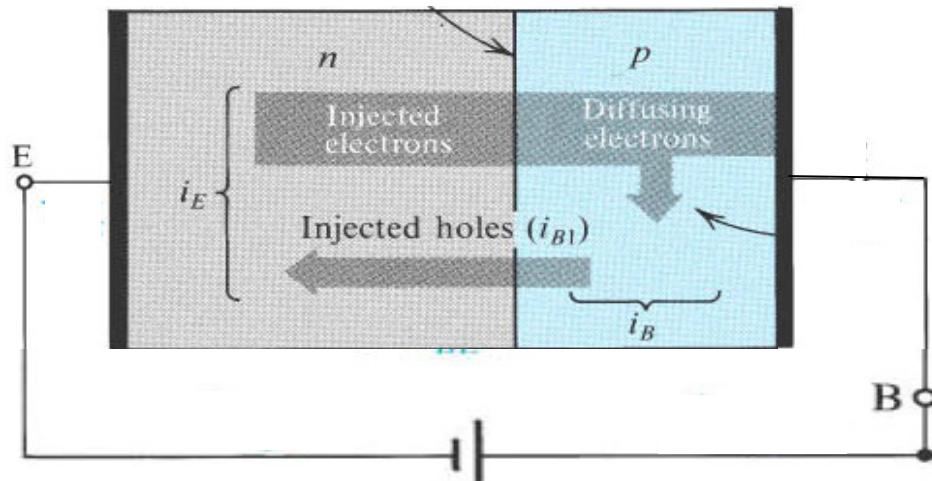
$$\frac{dQ}{dV_{BE}} = \tau_F \frac{dI_C}{dV_{BE}}$$

$$C_D = \tau_F \left(\frac{I_C}{\phi_t} \right)$$

$$\text{Ex: } \tau_F = 4 \times 10^{-10} \text{s}, \quad I_C = 1 \text{ mA} \quad \rightarrow \quad C_D = 16 \text{ pF}$$

Emitter efficiency

EB junction of npn transistor is an N⁺P junction. Electrons from the emitter are injected into the base, but holes from the base are also injected into the emitter. The ratio $\gamma = J_n/J_p$ is designated as emitter efficiency.



$$\int_E^B \frac{pJ_n dx}{D_n} = \int_E^B \frac{nJ_p dx}{D_p} \quad \xrightarrow{\text{Negligible recombination}}$$

$$\xrightarrow{\int_E^B ndx \cong \int_{\text{Emitter region}} N_{dE} dx; \quad \int_E^B pdx \cong \int_{\text{Base region}} N_{aB} dx}$$

Let us, once again, use

$$\frac{pJ_n}{D_n} - \frac{nJ_p}{D_p} = q \frac{d(pn)}{dx} \quad (1)$$

At the ohmic contacts E and B,

$$pn = n_i^2$$

Integrating (1) from E to B yields

$$\gamma = \frac{J_n}{J_p} = \frac{D_n}{D_p} \frac{\int_E^B ndx}{\int_E^B pdx} \quad \xrightarrow{\text{Low-level injection}}$$

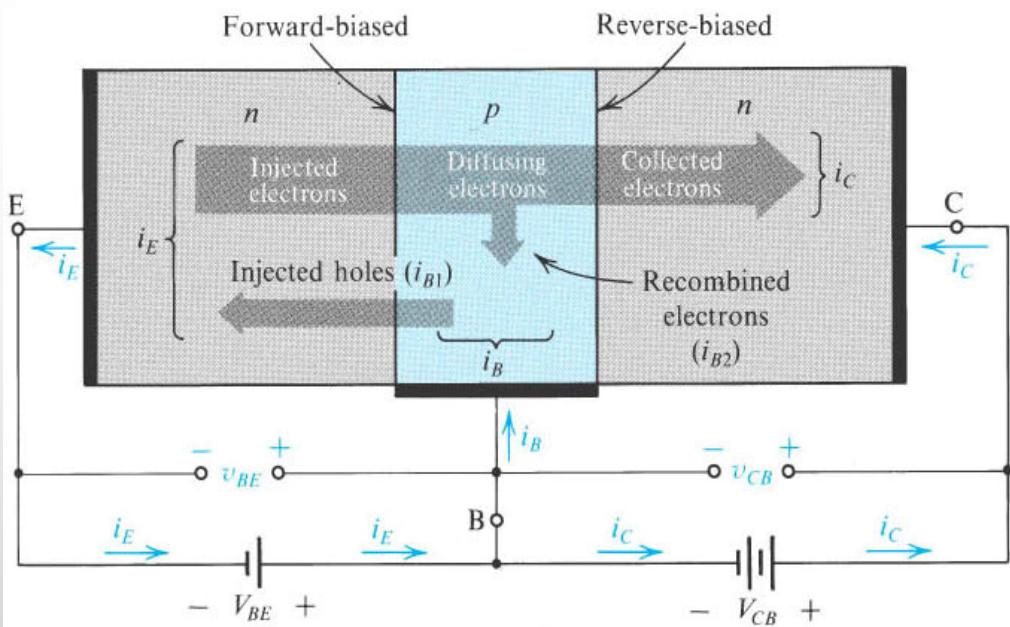
$$\int_E^B ndx \cong \int_{\text{Emitter region}} N_{dE} dx; \quad \int_E^B pdx \cong \int_{\text{Base region}} N_{aB} dx$$

$$\gamma = \frac{J_n}{J_p} = \frac{D_n}{D_p} \frac{\text{Emitter region}}{\int_{\text{Base region}} N_{aB} dx}$$

Emitter efficiency

If the recombination current is negligible, we have

$$I_p = \frac{I_n}{\gamma} = \frac{D_p}{D_n} \frac{\int_{\text{Base region}} N_{aB} dx}{\int_{\text{Emitter region}} N_{dE} dx} I_n \ll I_n$$

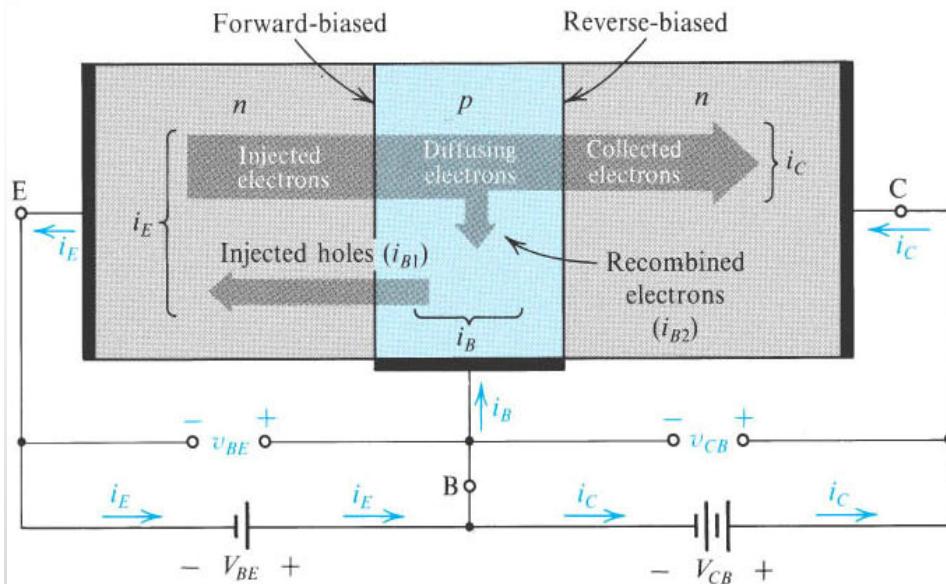


Base transport factor - α_T

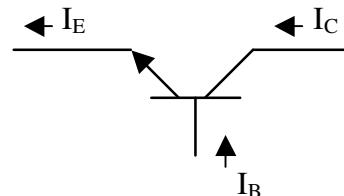
$$\alpha_T = \frac{I_{nC}}{I_{nE}} = \frac{1}{\cosh(W_B / L_n)} \cong 1 - \frac{W_B^2}{2L_n^2}$$

Example: $\frac{W_B}{L_n} = \frac{0.2\mu\text{m}}{20\mu\text{m}} \rightarrow \alpha_T \cong 1 - \frac{W_B^2}{2L_n^2} = 1 - \frac{1}{20000}$

The base current



Forward-active mode



$$I_B = I_{B1} + I_{B2} + I_{B3} \cong \frac{I_C}{\gamma} + (1 - \alpha_T) I_C + I_{B3}$$

The Base Current

I_{B1} : holes injected from the base region into the emitter region

I_{B2} : some electrons that diffuse across the base do not reach the collector, but on the way they recombine with holes. The missing holes must be replaced from the external circuit.

I_{B3} : recombination current in the EB junction (important at low current levels)

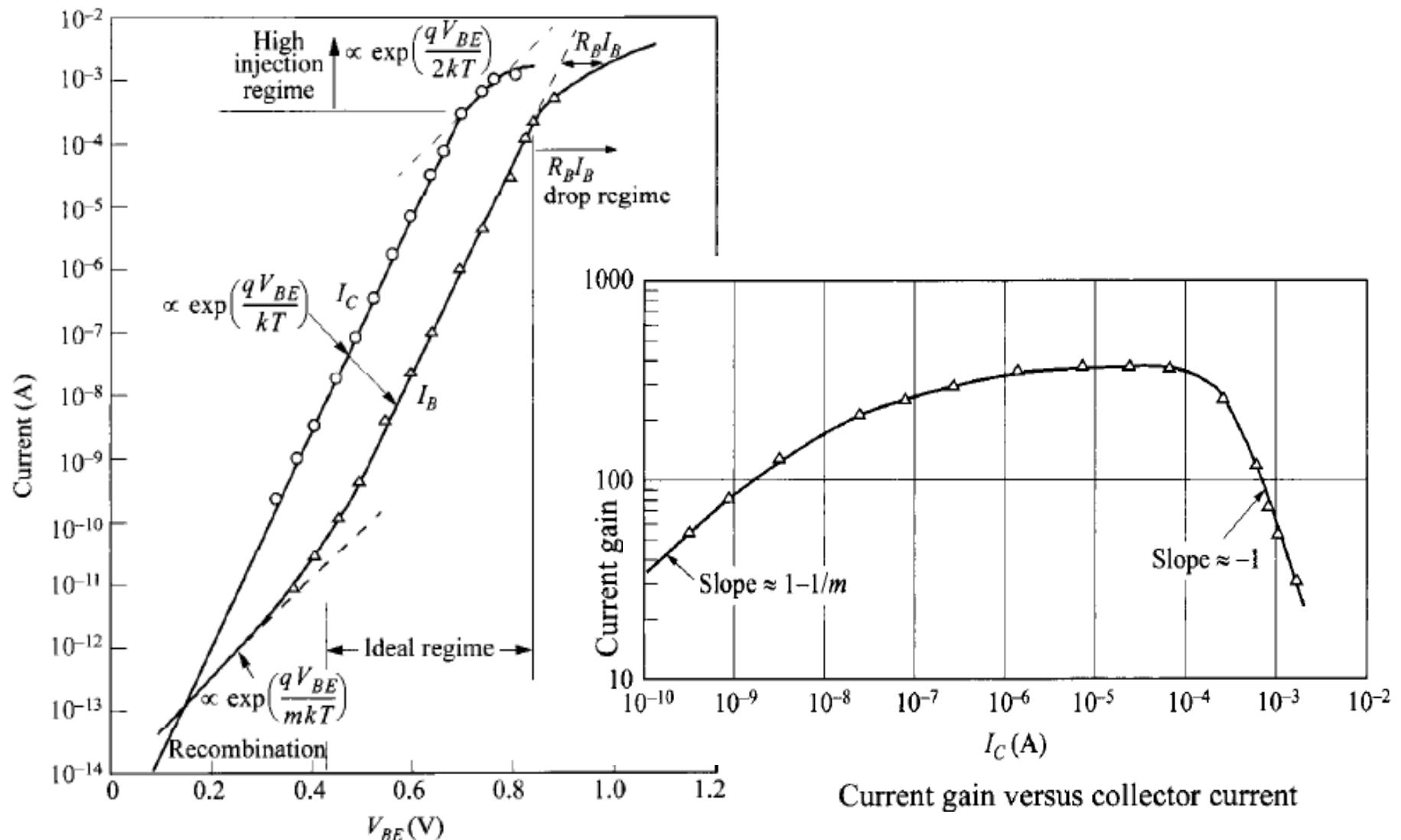
Usually $I_{B2} \ll I_{B1} \rightarrow I_B \cong I_{B1} + I_{B3}$

$$I_E = I_B + I_C$$

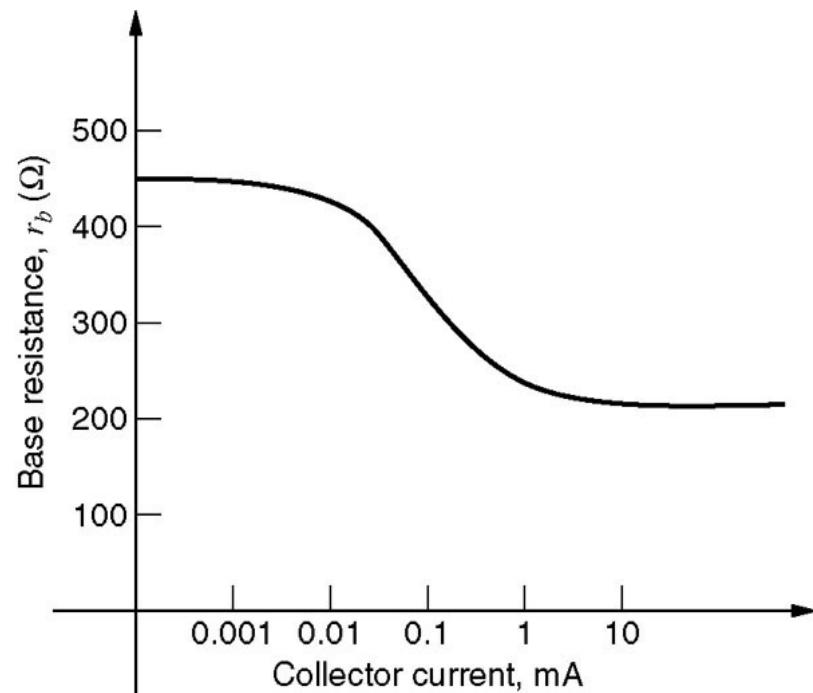
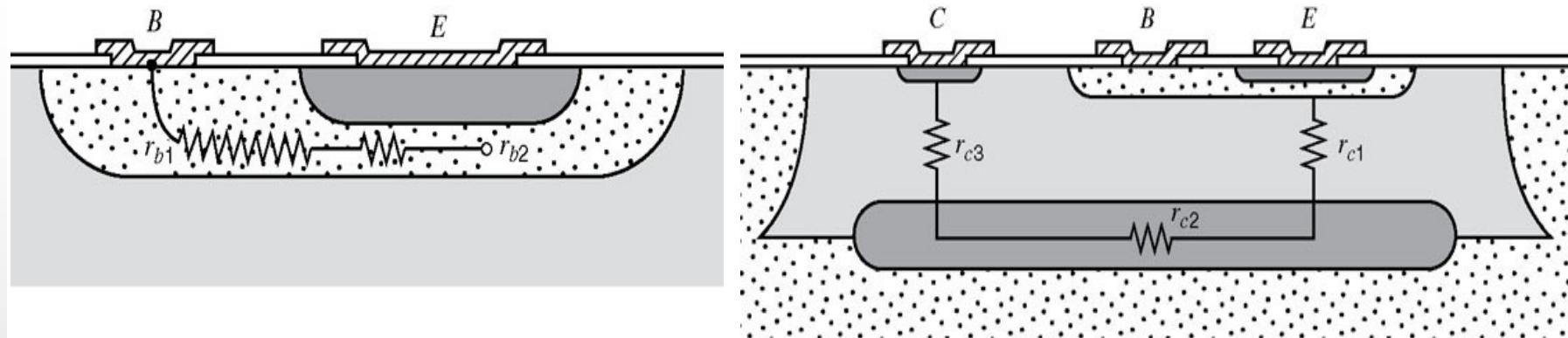
$$I_C = \alpha_F I_E$$

$$\beta_F = \frac{\alpha_F}{1 - \alpha_F}$$

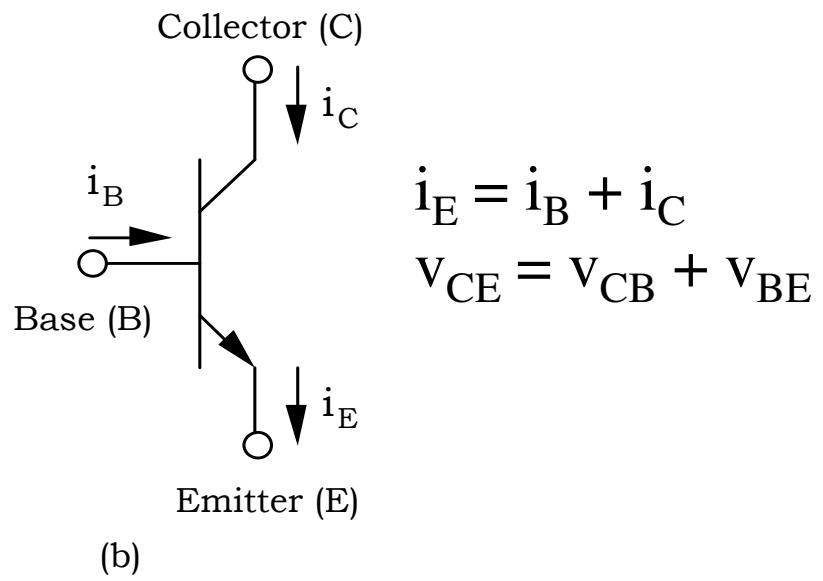
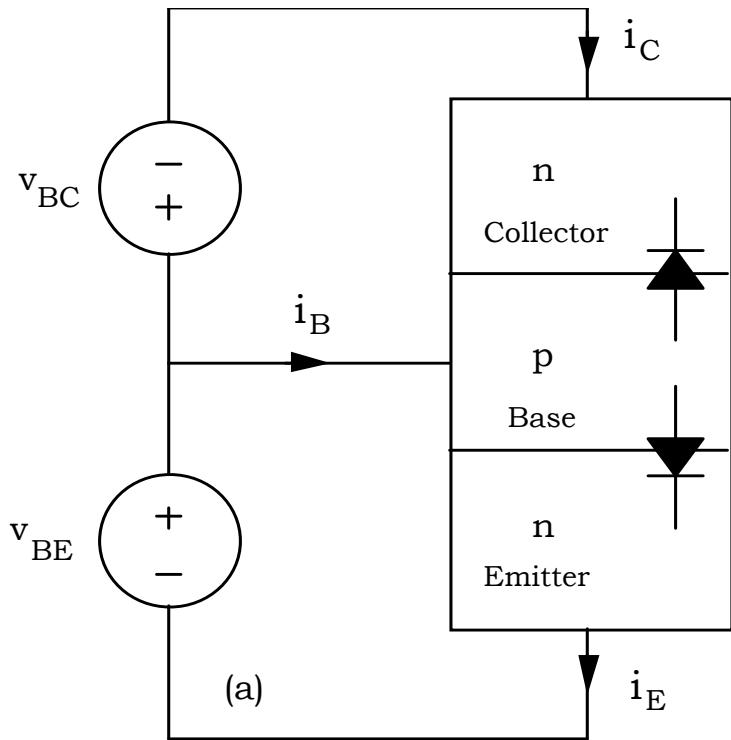
$$I_C = \beta_F I_B$$



Base and collector resistances

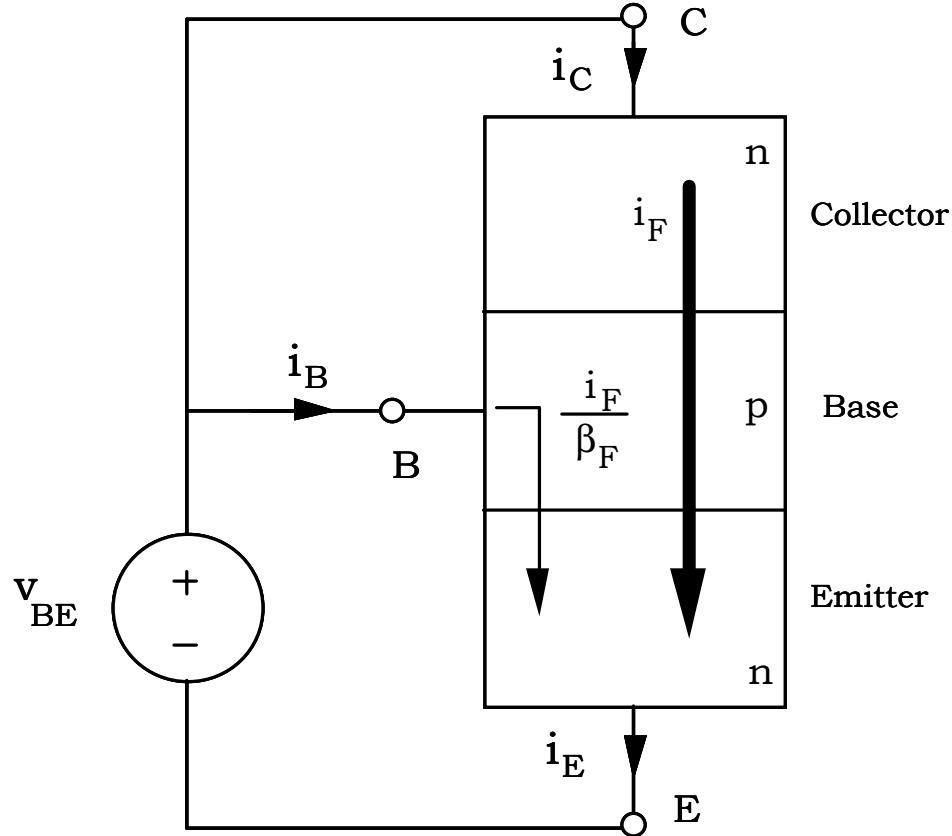


The (Ebers-Moll) Transport Model of the NPN Transistor



- (a) Idealized npn transistor structure for a general bias condition
- (b) Circuit symbol for the npn transistor

Forward Transport Current



NPN transistor with $V_{BE} > 0$ and $V_{BC} = 0$.

$$i_C = i_F = I_s \left[\exp\left(\frac{V_{BE}}{\phi_t}\right) - 1 \right]$$

$$i_B = \frac{i_F}{\beta_F} = \frac{I_s}{\beta_F} \left[\exp\left(\frac{V_{BE}}{\phi_t}\right) - 1 \right]$$

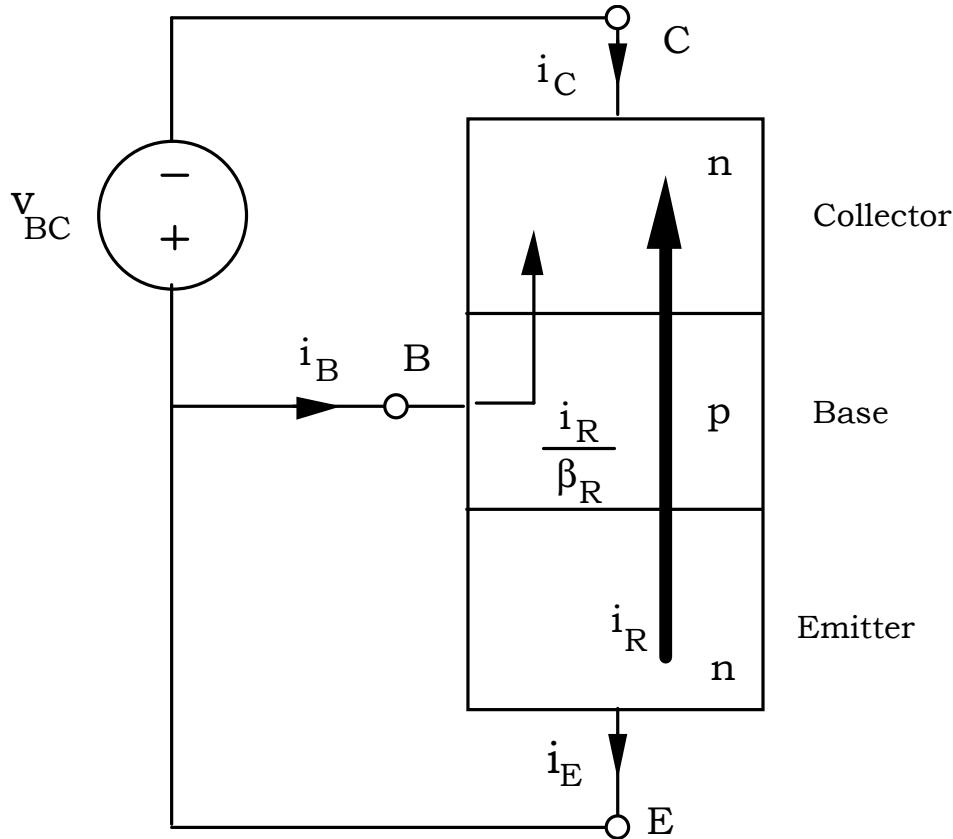
$$20 \leq \beta_F \leq 500 \text{ (typically)}$$

β_F : forward common-emitter current gain

$$i_E = i_C + i_B = \frac{\beta_F + 1}{\beta_F} i_C = \frac{i_C}{\alpha_F}$$

α_F : forward common-base current gain

Reverse Transport Current



Transistor with $V_{BC} > 0$ and $V_{BE} = 0$.

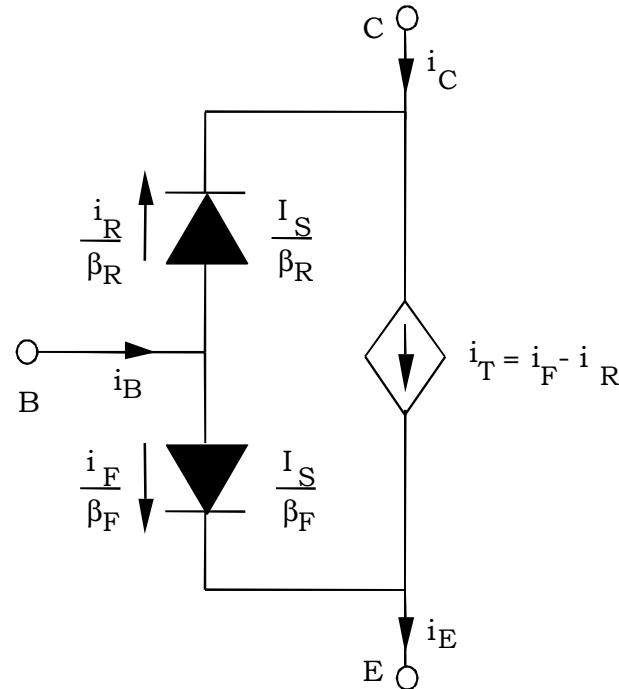
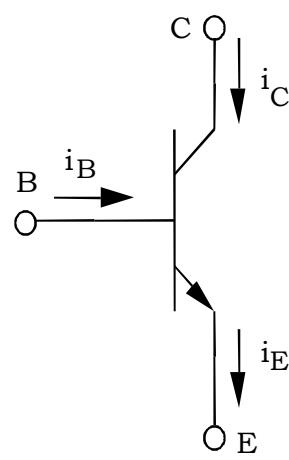
$$i_E = -i_R = -I_S \left[\exp\left(\frac{V_{BC}}{\phi_t}\right) - 1 \right]$$

$$i_B = \frac{i_R}{\beta_R} = \frac{-I_S}{\beta_R} \left[\exp\left(\frac{V_{BC}}{\phi_t}\right) - 1 \right]$$

$$0.02 \leq \beta_R \leq 20$$

$$i_C = \frac{i_E}{\alpha_R} = \frac{\beta_R + 1}{\beta_R} i_E$$

The Transport Model of the NPN Transistor



$$i_F = I_S \left[\exp(v_{BE} / \phi_t) - 1 \right]$$

$$i_R = I_S \left[\exp(v_{BC} / \phi_t) - 1 \right]$$

$$i_B = \frac{i_F}{\beta_F} + \frac{i_R}{\beta_R}$$

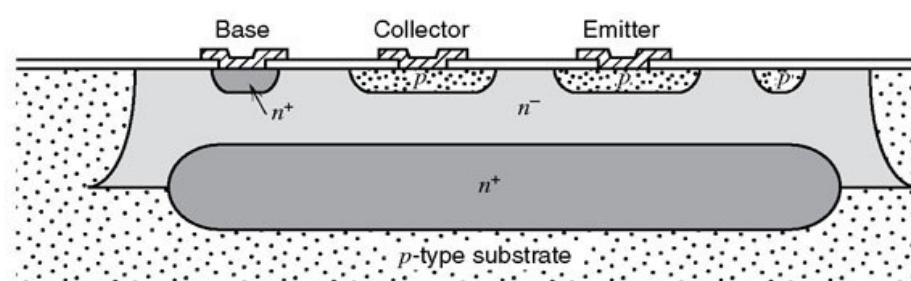
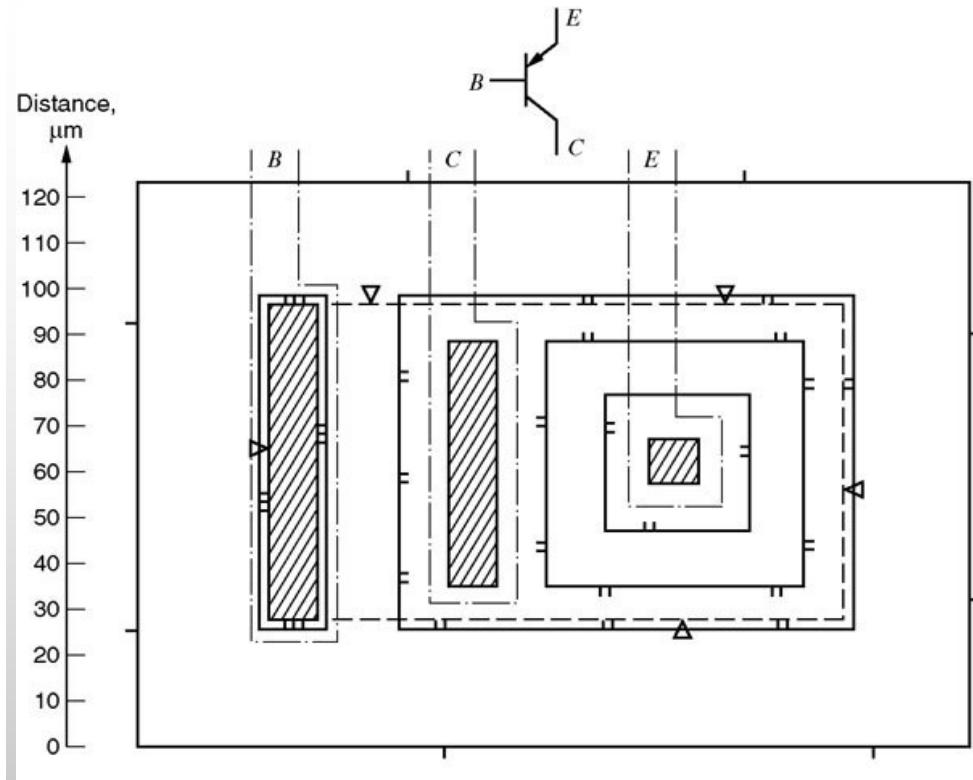
$$i_C = i_T - \frac{i_R}{\beta_R}$$

$$i_E = i_T + \frac{i_F}{\beta_F}$$

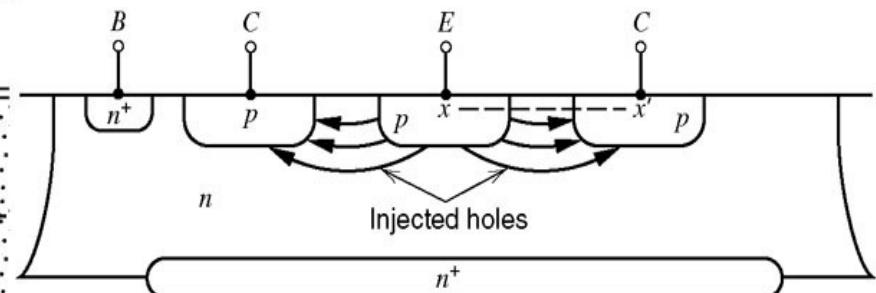
$$i_T = I_S \left[\exp(v_{BE} / \phi_t) - \exp(v_{BC} / \phi_t) \right]$$

Transport model equivalent circuit for the npn transistor

The lateral PNP transistor

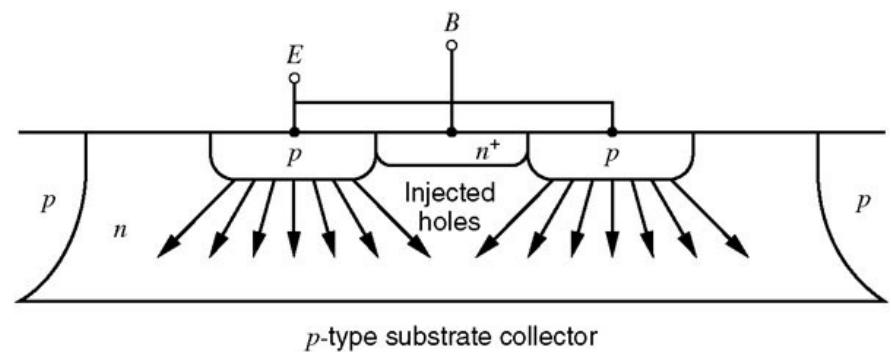
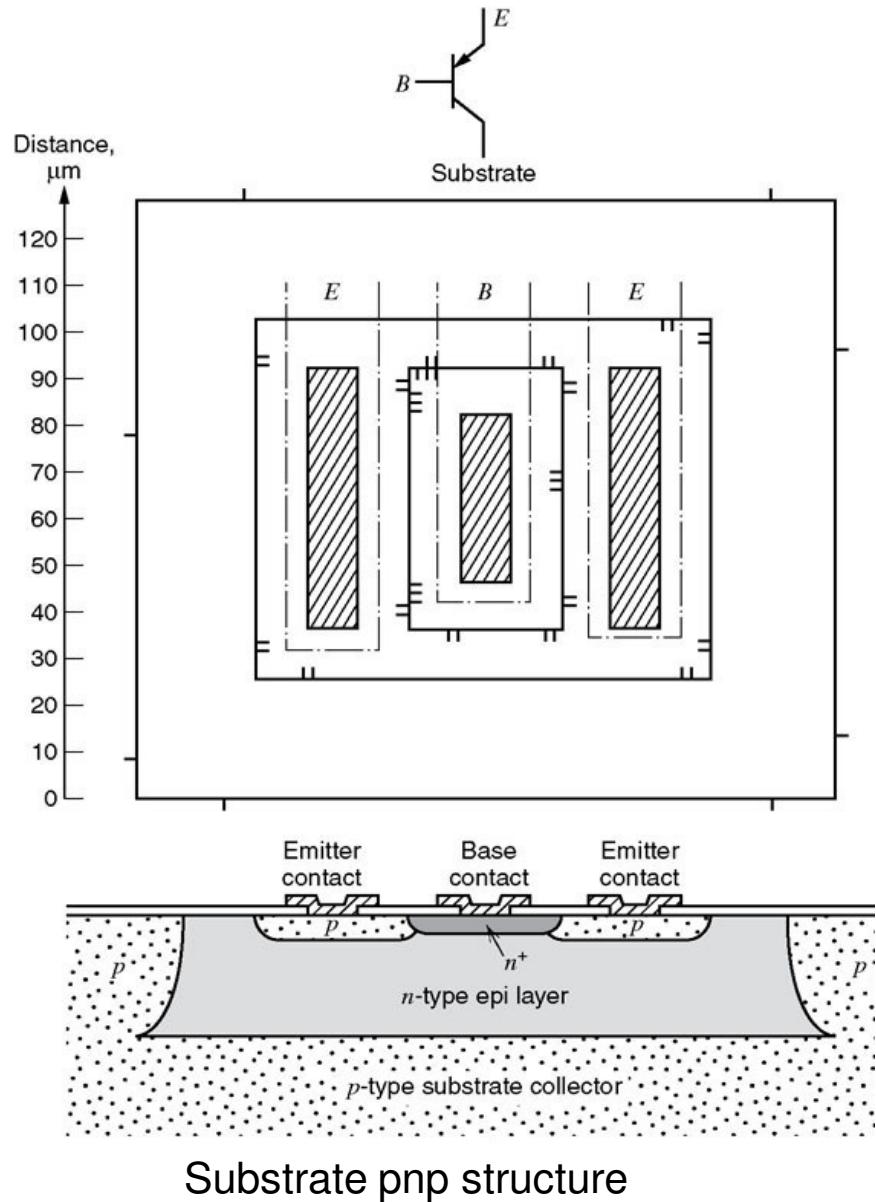


Lateral pnp structure



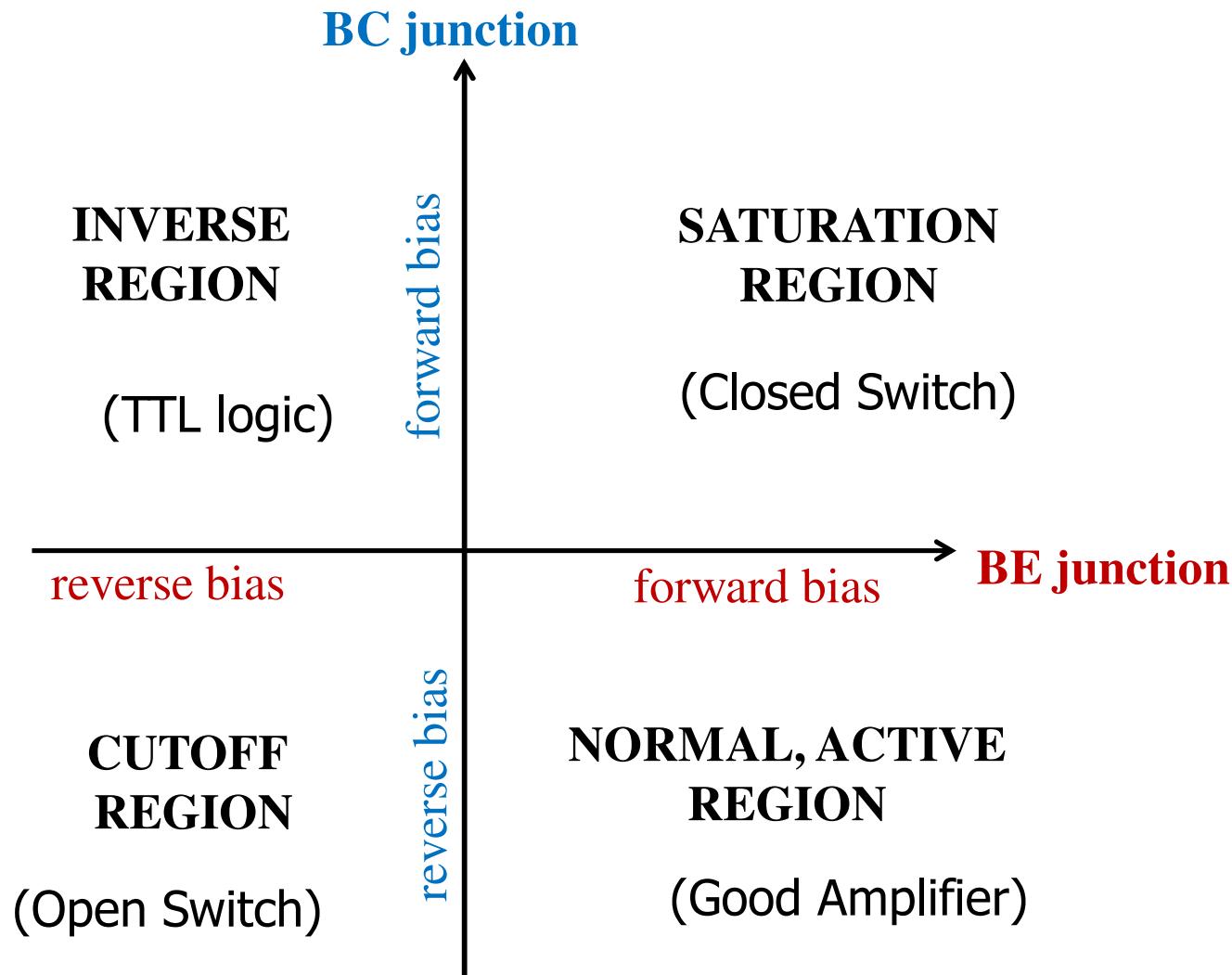
Minority carrier flow in
the lateral *pnp* transistor

The substrate PNP transistor

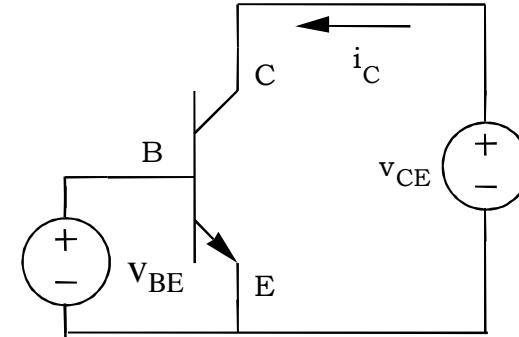
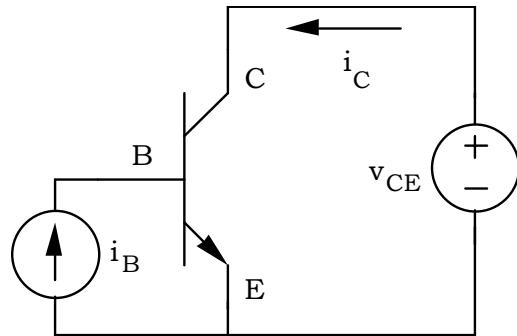


Minority carrier flow in the
substrate *pnp* transistor

Operating regions of the BJT

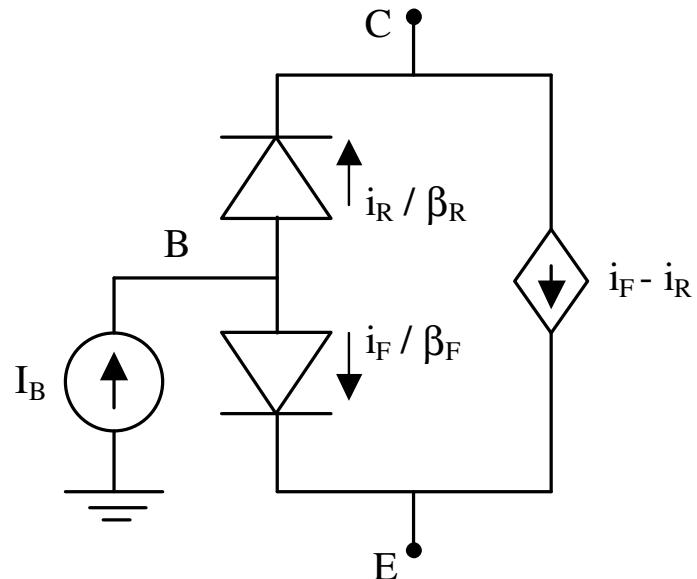


Graphical Representation of the BJT Characteristics



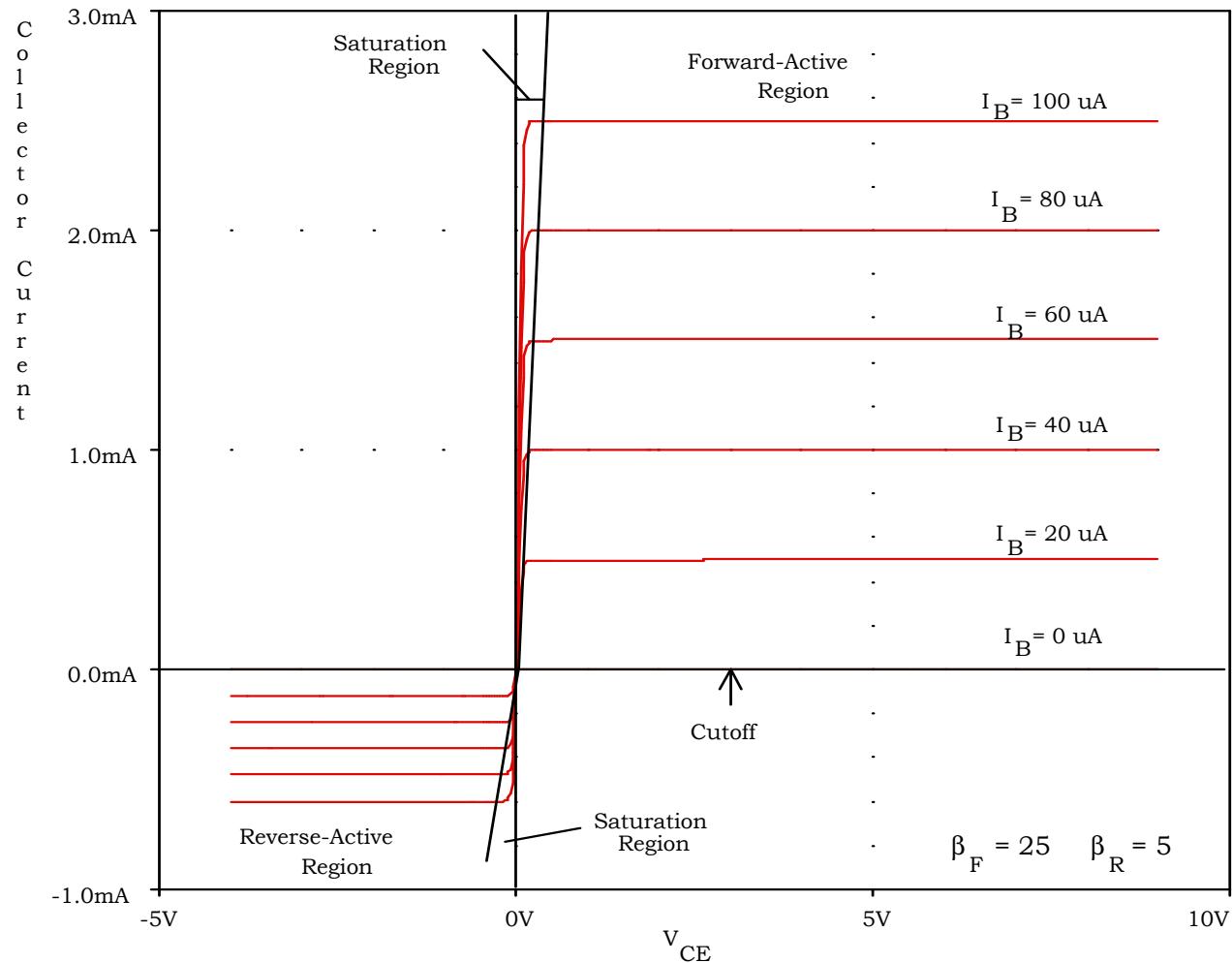
1st quadrant

Circuits for determining the common-emitter output characteristics



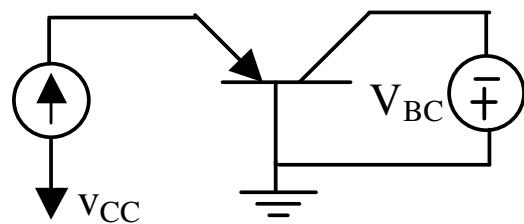
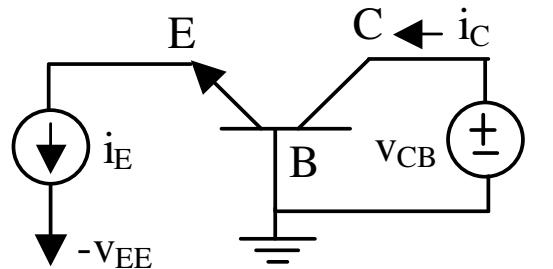
$$i_B = \frac{I_S}{\beta_F} \left[\exp\left(\frac{v_{BE}}{\phi_t}\right) - 1 \right] + \frac{I_S}{\beta_R} \left[\exp\left(\frac{v_{BC}}{\phi_t}\right) - 1 \right]$$

Graphical Representation of the BJT Characteristics

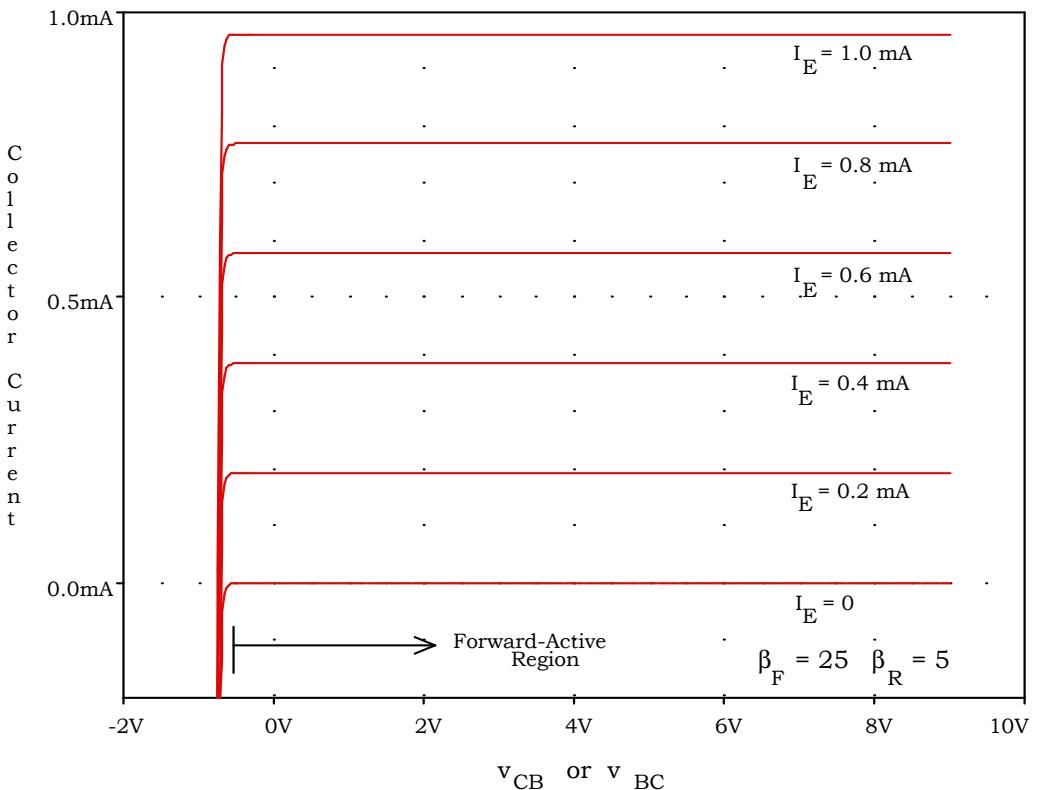
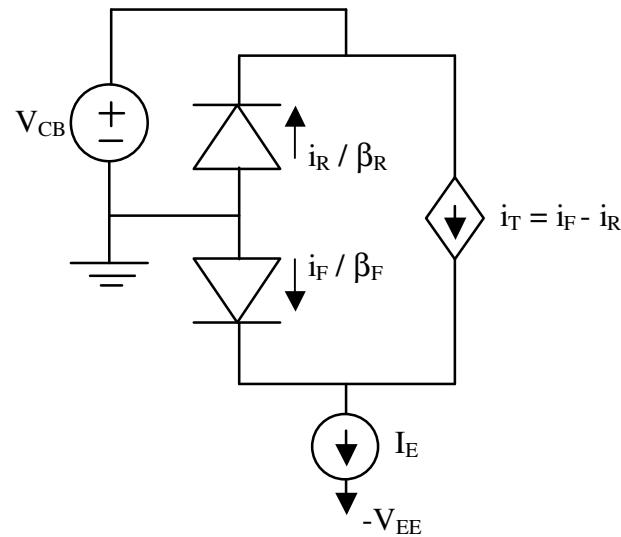


Common-emitter output characteristics for the BJT

Graphical Representation of the BJT Characteristics



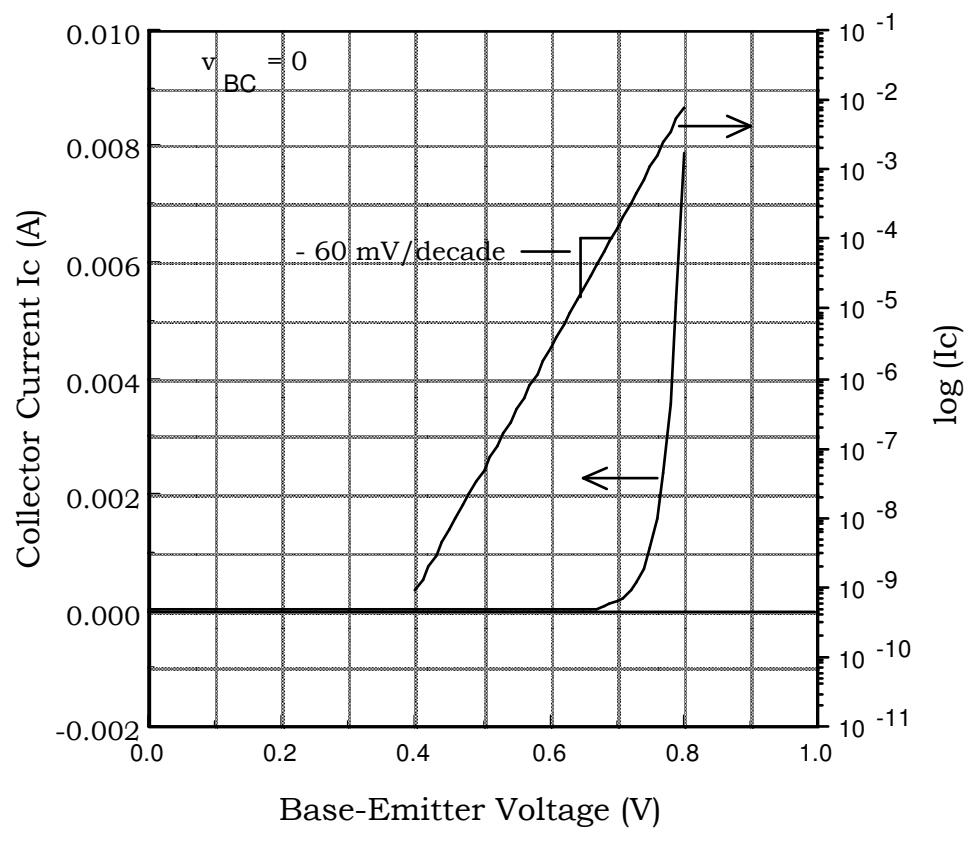
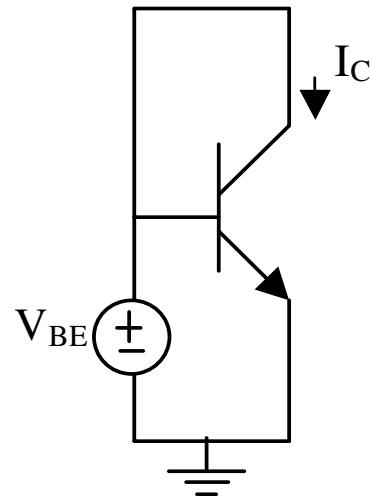
Circuit to determine common-base output characteristics



Common-base output characteristics for the BJT
 $(i_C$ vs. v_{CB} for npn BJT or i_C vs. v_{BC} for the pnp BJT)

Graphical Representation of the BJT Characteristics

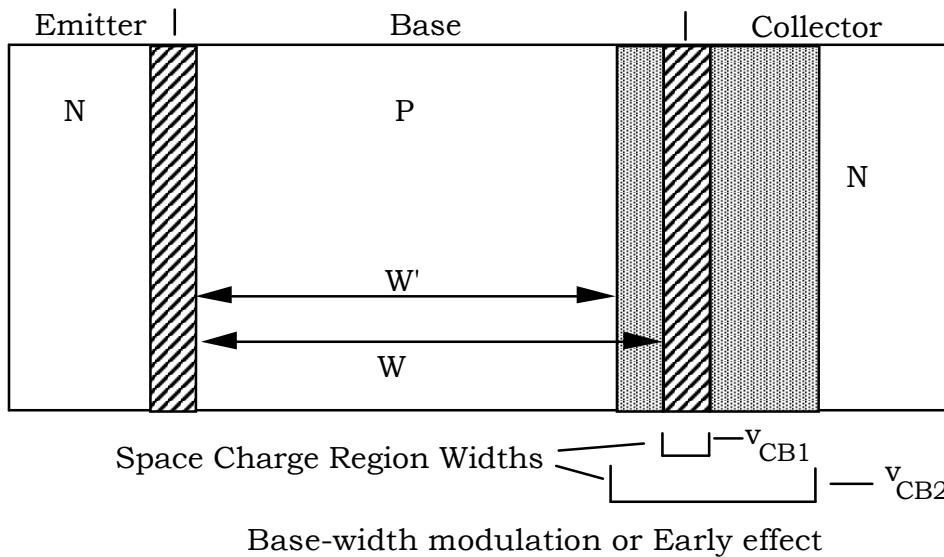
$$I_C = I_S \left[\exp\left(\frac{V_{BE}}{\phi_t}\right) - 1 \right]$$



BJT transfer characteristic in the forward-active region

$$\left. \frac{\Delta V_{BE}}{\Delta T} \right|_{I_C} \approx -2 \text{ mV/}^\circ\text{C} \rightarrow \text{Dependent of } I_C$$

The Early Effect



Active mode

$$i_C = I_S \exp(v_{BE} / \phi_t)$$

$$I_S = \frac{qAD_n n_i^2}{W_B}; \rightarrow I_S = \frac{qAD_n n_i^2}{N_{aB} W_B}$$

Uniform base
doping

Modulated by
 V_{CB} (or V_{CE})

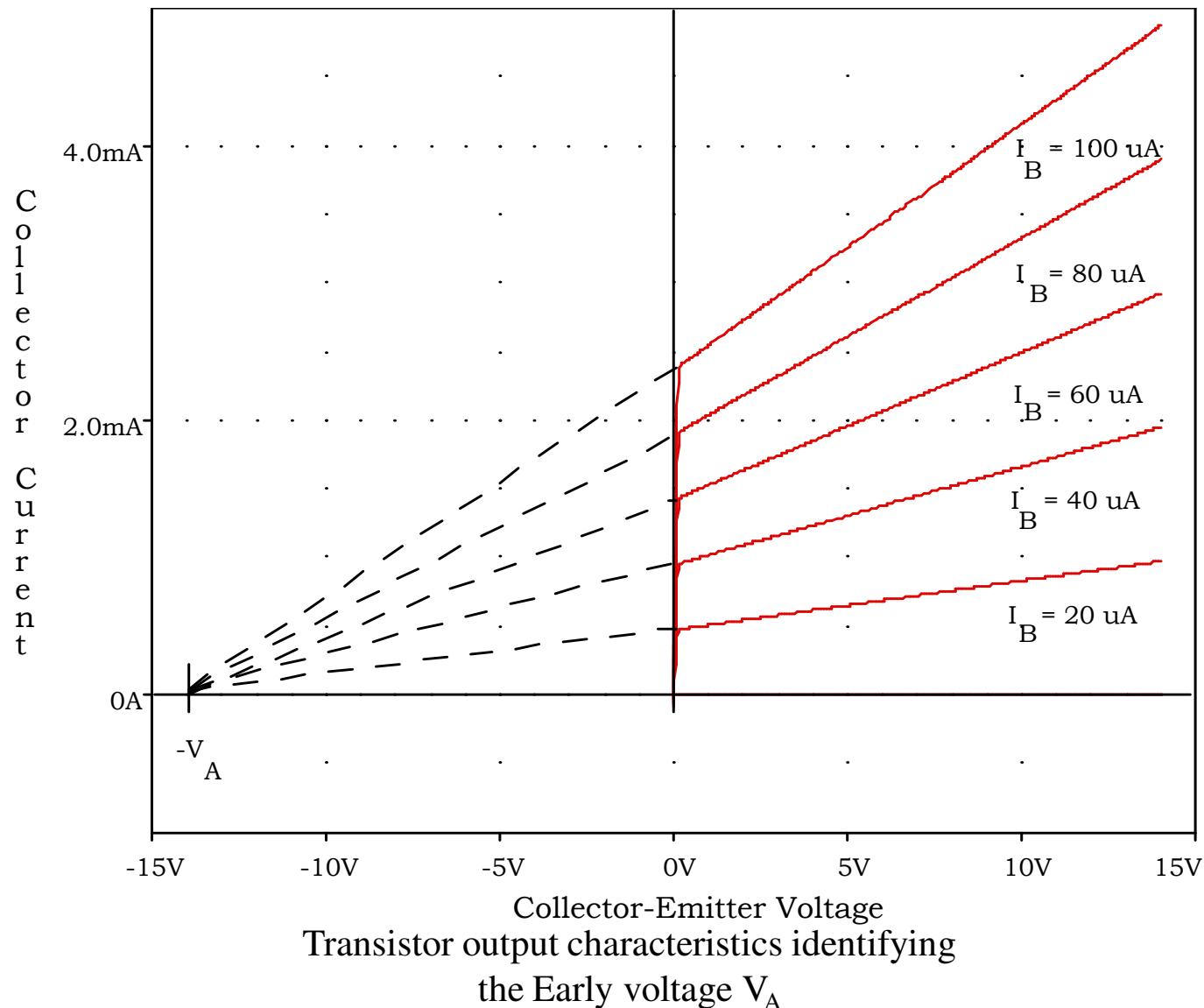
Using linear approximation for W_B

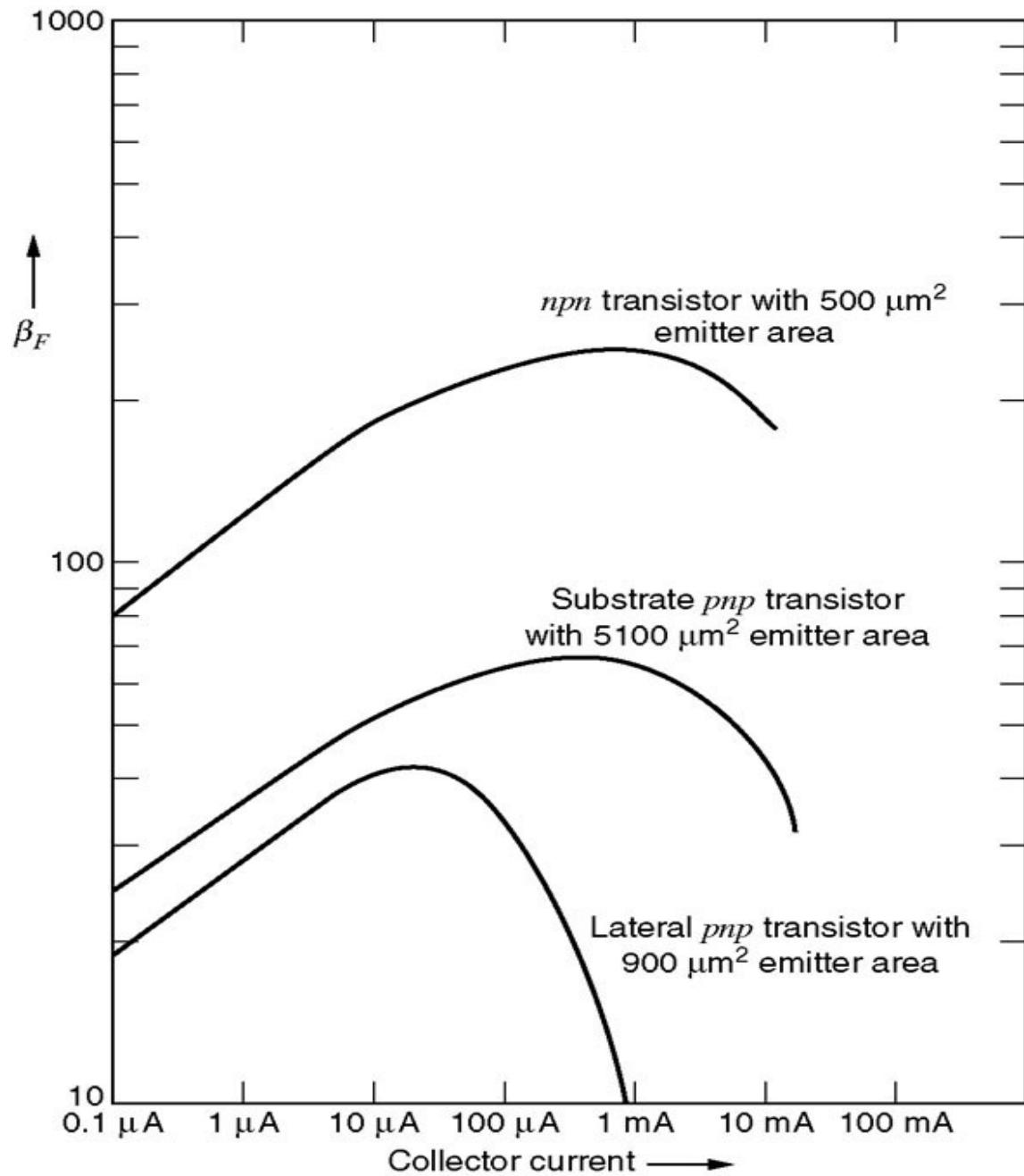
$$W_B = W_{B0} (1 - v_{CE} / V_A) \text{ yields}$$

$$i_C = \frac{I_{SO}}{(1 - v_{CE} / V_A)} \exp \left(\frac{v_{BE}}{\phi_t} \right) \cong I_{SO} \exp \left(\frac{v_{BE}}{\phi_t} \right) \left(1 + v_{CE} / V_A \right)$$

$v_{CE} / V_A \ll 1 \quad V_A : \text{Early voltage}$

The Early Voltage





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Parameter	Typical Value, 5- Ω -cm, 17- μm epi 44-V Device 5100 μm^2 Emitter Area	Typical Value, 1- Ω -cm, 10- μm epi 20-V Device 5100 μm^2 Emitter Area
β_F	50	30
β_R	4	2
V_A	50 V	30 V
η	5×10^{-4}	9×10^{-4}
I_S	10^{-14} A	10^{-14} A
I_{CO}	2×10^{-10} A	2×10^{-10} A
BV_{CEO}	60 V	30 V
BV_{CBO}	90 V	50 V
BV_{EBO}	7 V or 90 V	7 V or 50 V
τ_F	20 ns	14 ns
τ_R	2000 ns	1000 ns
β_0	50	30
r_b	150 Ω	50 Ω
r_c	50 Ω	50 Ω
r_{ex}	2 Ω	2 Ω
Base-emitter junction	$\begin{cases} C_{je0} & 0.5 \text{ pF} \\ \psi_{0e} & 0.55 \text{ V} \\ n_e & 0.5 \end{cases}$	$\begin{cases} 1 \text{ pF} \\ 0.58 \text{ V} \\ 0.5 \end{cases}$
Base-collector junction	$\begin{cases} C_{\mu 0} & 2 \text{ pF} \\ \psi_{0c} & 0.52 \text{ V} \\ n_c & 0.5 \end{cases}$	$\begin{cases} 3 \text{ pF} \\ 0.58 \text{ V} \\ 0.5 \end{cases}$

Parameter	Vertical <i>npn</i> Transistor with $2 \mu\text{m}^2$	Lateral <i>pnp</i> Transistor with $2 \mu\text{m}^2$
	Emitter Area	Emitter Area
β_F	120	50
β_R	2	3
V_A	35 V	30 V
I_S	$6 \times 10^{-18}\text{A}$	$6 \times 10^{-18}\text{A}$
I_{CO}	1 pA	1 pA
BV_{CEO}	8 V	14 V
BV_{CBO}	18 V	18 V
BV_{EBO}	6 V	18 V
τ_F	10 ps	650 ps
τ_R	5 ns	5 ns
r_b	400Ω	200Ω
r_c	100Ω	20Ω
r_{ex}	40Ω	10Ω
C_{je0}	5 fF	14 fF
ψ_{0e}	0.8 V	0.7 V
n_e	0.4	0.5
$C_{\mu 0}$	5 fF	15 fF
ψ_{0c}	0.6 V	0.6 V
n_c	0.33	0.33
C_{cs0} (C_{bs0})	20 fF	40 fF
ψ_{0s}	0.6 V	0.6 V
n_s	0.33	0.4